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On the Ground and in the Sky: A Tutorial on Radio Localization in Ground-Air-Space Networks

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Abstract—The inherent limitations in scaling up ground infrastructure for future wireless networks, combined with decreasing operational costs of aerial and space networks, are driving considerable research interest in multisegment ground-air-space (GAS) networks. In GAS networks, where ground and aerial users share network resources, ubiquitous and accurate user localization becomes indispensable, not only as an end-user service but also as an enabler for location-aware communications. This breaks the convention of having localization as a byproduct in networks primarily designed for communications. To address these imperative localization needs, the design and utilization of ground, aerial, and space anchors require thorough investigation. In this tutorial, we provide an in-depth systemic analysis of the radio localization problem in GAS networks, considering ground and aerial users as targets to be localized. Starting from a survey of the most relevant works, we then define the key characteristics of anchors and targets in GAS networks. Subsequently, we detail localization fundamentals in GAS networks, considering 3D positions, orientations, and velocities. Afterward, we thoroughly analyze radio localization systems in GAS networks, detailing the system model, design aspects, and considerations for each of the three GAS anchors. Preliminary results are presented to provide a quantifiable perspective on key design aspects in GAS-based localization scenarios. We then identify the vital roles 6G enablers are expected to play in radio localization in GAS networks.

Index Terms—5G, 6G, aerial networks, aerial targets, AI, cell-free, ground aerial space networks, JCAS, localization, LEO satellites, massive MIMO, non-terrestrial networks, RIS, THz, UAVs.

I. INTRODUCTION

O in wireless systems has shifted from being an optional feature to becoming an imperative functionality that drives the innovation of new wireless applications. Location information serves as the essence of many technological and socially

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impactful applications, in the areas of mobile devices, Internet of things (IoT), autonomous (aerial) vehicles, precision agriculture, as well as live augmented and virtual reality experiences. In applications such as autonomous vehicles [1], assisted living IoT [2], unmanned aerial vehicle (UAV) traffic management [3], and the vast majority of IoT applications [4]–[6], the dependency on location information is pivotal. Any error in the location information could lead to serious or even life-threatening consequences. For instance, imagine losing the track position in a self-driving car or having the global positioning system (GPS) signal jammed during a commercial flight journey or a UAV mission.

While localization in terrestrial networks is well incorporated in the different generations of cellular networks, from the first generation (1G) to the fifth generation (5G) [17], the integration of non-terrestrial networks (NTNs) in current 5G and future sixth generation (6G) wireless networks, necessitate a paradigm shift in localization systems, promoting novel ground-aerial-space (GAS) multisegment localization systems, in which terrestrial and non-terrestrial segments operate seamlessly, harnessing both horizontal and vertical dimensions. On the ground, current 5G wireless systems handle the increasing connectivity and spectral efficiency demands by employing base stations with a massive number of antennas [18] and exploring new spectrum bands, such as millimeterwave (mmWave) and Terahertz bands [19]. This combination of a large number of antennas and large radio bandwidth offers exceptionally sharp angle and time measurements, enabling unprecedentedly accurate localization. Departing from the ground upward, the non-terrestrial aerial and space segments follow, which include low-altitude platforms (LAPs), high altitude platforms (HAPs), and satellite networks. Wireless networks in the non-terrestrial segments, which are commonly known as NTNs, are included in recent study items of 3rd generation partnership project (3GPP) [20] as well as in 3GPP Release 17 and Release 18 packages [21]. Recent developments of aerial/space technologies coupled with reduced manufacturing costs have enabled various NTN-based advanced applications focusing on providing continuous, ubiquitous, and high-capacity connectivity across the globe [7]. Employing NTNs for connectivity naturally unlocks the potential of also providing radio frequency (RF)-based localization as a joint service by design. In this tutorial, we provide a systematic analysis of the radio localization system design in the different segments of the GAS networks, concentrating on the scope detailed in the following.

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Ref.	Year	Sec. IV Ground Anchors		Sec. V Aerial Anchors		Sec. VI Space Anchors		Sec. VII Towards 6G GAS Localization					
Kei.		Ground targets	Aerial targets	Ground targets	Aerial targets	Ground targets	Aerial targets	RIS	JCAS	ΑI	CF	THz	KVIs
[7]	2022		ð	✓					ð				
[8]	2022	д		✓									
[9]	2022	✓	ð	✓					✓				
[10]	2022	✓		ð					✓				
[11]	2022	✓		ð				✓				✓	
[12]	2022			✓	ð					ð			
[13]	2022	✓							д	д			
[14]	2023	✓							ð				
[15]	2023	д				✓			д				
[16]	2023	✓	д			д		д	д	✓		д	
This	work	✓	✓	✓	✓	✓	✓	✓	√	✓	✓	✓	√

TABLE I: Topic-wise comparison with existing surveys and tutorials with $\sqrt{\text{and } \partial}$ denoting covered and partially covered, respectively.

A. Scope

In this work, we focus on the localization system design within the following scope.

- Radio technologies: The theory of localization is well
 mature from the geometry-based perspective. However,
 key challenges lie in the ways the range and direction estimates are realized and combined, which can be done via
 RF, computer vision, and inertia. The main scope of this
 work is RF-based localization in wide-area networks, and
 hence, it excludes short-range technologies such as nearfield communications, ultra-wideband, and Bluetooth.
- GAS anchors: Satellite-based localization, e.g., global navigation satellite system (GNSS) is the most popular way to localize a device, where GNSS constellations are generally operated in medium Earth orbit (MEO) and geostationary orbit (GEO). In this paper, we look beyond GNSS and investigate localization using ground, aerial, and low Earth orbit (LEO) satellites.
- Ground and aerial targets/users¹: Given that satellites are launched to a predefined orbit, in this work, we focus on the ground and aerial target that emits and/or receives RF signals, and as such, passive devices are out of the scope.
- Outdoor environments: Over the past decade non-GNSS-based localization systems were mainly focused on indoor environments. In this paper, we address the immense demand for GNSS alternative outdoor localization methods, breaking from the confines of indoor short-range localization systems.

B. Related Work

The vast majority of related tutorials and surveys in the literature addressed either GAS networks mainly from wireless communications perspective [7,8], or localization systems mainly from the ground segment perspective [9]–[16,22]. A detailed topic-wise comparison between this tutorial and recent relevant tutorials and surveys is presented in Table I. The table shows that the majority of recent surveys and tutorials in the literature mainly focus on terrestrial localization systems, highlighting the gap regarding tutorials and surveys on non-terrestrial aerial and space localization systems. Broadly speaking, the technical content of related surveys and tutorials

can be categorized into terrestrial localization systems, 6Genabled localization systems, and non-terrestrial localization systems.

Surveys and tutorials studying terrestrial localization systems, which address the localization requirements of different ground users, e.g., people, vehicles, and IoT nodes, can be found in [13,14,22]. The authors of [22] presented an outlook on autonomous vehicle localization methods, along with a comparison and assessment of respective localization methods. A survey on the state-of-the-art in device-based localization and device-free sensing using mmWave communication and radar devices is provided in [13]. Active simultaneous localization and mapping (SLAM) approaches were surveyed in [14]. Despite their high relevancy, the relatively short-range ground-based localization systems become rather limited when considering diverse demands not only from the ground but also from the aerial user equipments (UEs). Starting from 2022, there was a noticeable shift in the scope of localization surveys and tutorials towards 6G enablers. In [9], the advancement of localization ability in 6G networks was emphasized, and a unified study on integrated localization and communication was conducted in [10]. The authors of [23] discussed five facets of 6G, one of which focused on radio localization and sensing. A tutorial on terahertz (THz) and mmWave localization systems was presented in [11]. The added value of reconfigurable intelligent surfaces (RISs) in 6G localization systems was discussed in [24]-[26], detailing use cases, challenges, and the road ahead. However, the localization service in the aforementioned works on 6G was mainly tailored for ground targets by using ground anchors, such as cellular base stations (BSs).

Regarding non-terrestrial localization systems, in [7] and [8], the evolution of NTNs in terms of infrastructure and network architecture, transitioning from 5G to 6G, are comprehensively discussed, emphasizing recent advances, scenarios, and envisioned role of NTNs in 5G and 6G networks. The localization accuracy obtained using aerial anchors in IoT networks was succinctly highlighted in [7] without elaborating on fundamental aspects of non-terrestrial localization systems. Moreover, localization aspects were not fully covered for ground and space segments. In [15], the authors presented a survey on state-of-the-art localization techniques for IoT applications using ground and space anchors. In particular, they discussed the localization performance in IoT networks when

¹Throughout this work, we use *target* and *user* interchangeably to refer to the target to be localized.

using solutions based on terrestrial anchors, GNSS anchors, and LEO satellite anchors. Localization in the air segment requires more degrees of freedom (DOFs) to improve the localization accuracy, especially in the vertical planes where aerial nodes fly. The authors of [12] provided a survey on UAV and swarm UAVs localization. As an important evolution, space anchors enabled by LEOs were briefly discussed in [16] for positioning ground targets without comprehensive emphasis on aerial anchors and targets. Because of the steadily increasing density of aerial vehicles in the sky, aerial target localization has become more prevailing than ever before [8,10,16].

Finally, in addition to the fact that most existing surveys and tutorials focus on terrestrial localization systems, from a system design point of view, only selected works detail mathematical modeling for the discussed localization technologies, such as [10,11,16]. However, these works do not address localization in NTNs, stressing the need for tutorials to comprehensively investigate the fundamentals and practical implementations of various anchors and targets in integrated GAS networks. To our best knowledge, there is no such thorough study that comprehensively investigates the localization functionality using GAS networks. Therefore, this work, which presents relevant mathematical modeling and formulation, will facilitate the research community to better understand the localization system design with different GAS segments and thus deal with on-demand localization services realized by the most superior and suitable methods.

C. Contribution

Considering the tremendous potential to realize precise localization services using modern ground, aerial, and space anchors, there is an urgent need to provide comprehensive tutorials on how to fully utilize the advantages of such GAS segments. Unlike existing works in the literature, this tutorial is distinguished by introducing localization systems in GAS networks, detailing system modeling, key design aspects, main considerations for system optimization, as well as the corresponding localization algorithms used. In particular, this tutorial aims to introduce the design foundations and fundamental aspects of localization systems in GAS network. We put a spotlight on RF localization in all three segments in GAS networks by elaborating on key challenges to be addressed and opportunities to be leveraged in future deployments. Furthermore, we detail design aspects and considerations of integrating the main 6G enabler, namely RISs, joint communication and sensing (JCAS), artificial intelligence (AI), cellfree (CF), and THz, in GAS localization systems. Finally, we discuss localization key value indicators (KVIs) in the context of GAS localization systems, extending the perspective of the conventional localization key performance indicators (KPIs) to value-driven indicators.

D. Organization

This tutorial paper introduces the fundamentals of localization systems in GAS networks. Unlike existing tutorials, we detail the system modeling of each segment in GAS networks

TABLE II: The overall structure of the tutorial.

I. Introduction							
A. Scope	B. Related Work						
C. Contribution	D. Organization						
II. Beyond GNSS: GAS Anchors and Targets							
A. GNSS Limitations	B. Characteristics of GAS Anchors						
C. Characteristics of Ground and Aerial Targets	D. Localization Scenarios in GAS						
III. Localization Fund	damentals						
A. System Geometry in 3D	B. Localization Measurables						
C. Snapshot Localization Methods	D. From Snapshot to Tracking						
E. Key Performance Metrics	E. Key Performance Metrics						
IV. Localization with Ground Anchors (5G)							
A. System Model	B. Ground Targets: Design Aspects						
C. Aerial Targets: Special Considerations	D. Key Takeaways						
V. Localization with Aerial Anchors							
A. System Model	B. Ground Targets: Design Aspects						
C. Aerial Targets: Special Considerations	D. Key Takeaways						
VI. Localization with Sp	pace Anchors						
A. System Model	B. Ground Targets: Design Aspects						
C. Aerial Targets: Special Considerations	D. Key Takeaways						
VII. Research Directions Towards 6G GAS Localization							
A. RIS	B. JCAS						
C. AI-Empowered Localization	D. Cell-Free Paradigm						
E. THz Band	F. 6G KVIs						
G. Key Takeaways	G. Key Takeaways						
VIII. Conclusion							

and the key design aspects and considerations with respect to both ground and non-terrestrial targets. Moreover, we position GAS segments into the perspective of the future that 6G networks enable, elaborating on the role and advantages to be gained from each of the enablers. In particular, the rest of the paper is organized as follows. Section II details the rationale that urges for GNSS localization alternatives and introduces the characteristics of GAS anchors and targets. In Section III, we present the RF-based localization fundamentals, and we put a special emphasis on the 3D space associated with GAS networks. Sections IV, V, and VI, respectively, present ground, aerial, and space anchors in GAS networks. In each one of these sections, we detail the corresponding system model, design aspects, and special considerations for ground and aerial target localization. Section VII discusses the integration of 6G enablers in GAS networks from a localization point of view and presents localization KVIs in the context of GAS localization systems. Key takeaways are summarized at the end of Section IV-VII. Finally, we conclude this tutorial in Section VIII. The overall structure of the tutorial is presented in Table II.

The *acronyms* list is presented in Table III. *Notation*: Italic, simple bold, and capital bold letters represent scalars, vectors, and matrices, respectively. We use $(a_1, a_2, ...)$ to represent a sequence and $\mathbf{a} = [a_1, a_2, ...]^{\mathsf{T}}$ to represent a column vector, with $[.]^{\mathsf{T}}$ being the transpose operator.

II. BEYOND GNSS: GAS ANCHORS AND TARGETS

This section presents the main limitations of GNSS-based localization systems. Subsequently, we introduce the main

TABLE III: List of acronyms

Acronym	Description	Acronym	Description	
1G	1st generation of wireless telecommunication	3GPP	3rd generation partnership project	
5G	5th generation of wireless telecommunication	6G	6th generation of wireless telecommunication	
A2A	Air-to-air	A2G	Air-to-ground	
ADC	Analog-to-digital converter	ADS-B	Automatic dependent surveillance-broadcast	
AI	Artificial intelligence	AP	Access point	
AOA	Angle-of-arrival	AOD	Angle-of-departure	
ATM	Air traffic management	BS	Base station	
C&C	Command and control	CDF	Cumulative distribution function	
CDMA	code-division multiple access	CF	Cell-free	
CF-mMIMO	Cell-free massive MIMO	CFO	Carrier frequency offset	
CRLB	Cramér-Rao lower bound	CSI	Channel state information	
DOF	Degree(s) of freedom	G2A	Ground-to-air	
G2G	Ground-to-ground	GAS	Ground-aerial-space	
GCS	Global coordinate system	GDOP	Geometric dilution of precision	
GEO	Geostationary orbit	GNSS	Global navigation satellite system	
GPS	Global positioning system	HAP	High altitude platform	
IMU	Inertial measurement unit	IOC	Intersection of circles	
IoT	Internet of things	JCAS	Joint communication and sensing	
KF	Kalman filter	LAP	Low-altitude platform	
LCS	Local coordinate system		Low Earth orbit	
LOS	Line-of-sight	LR-FHSS	Long-range frequency hopping spread spectrum	
MEO			Multiple-input multiple-output	
MISO	Multiple-input single-output	ML	Machine learning	
MLAT	Multilateration	mmWave	Millimeter-wave	
NLOS	Non-line-of-sight	NTN	Non-terrestrial network	
OCXO	Oven controlled crystal oscillator	OISL	Optical inter-satellite link	
OFDM	Orthogonal frequency division multiplexing	PNT	Positioning, navigation, and timing	
PRN	Pseudo-random noise	PRS	Positioning reference signal	
RF	Radio frequency	RIS	Reconfigurable intelligent surface	
RMSE	Root-mean-square error	RSS	Received signal strength	
RTT	Round-trip time	S2A	Space-to-air	
S2G	Space-to-ground	SIMO	Single-input multiple-output	
SISO	Single-input single-output	SLAM	Simultaneous localization and mapping	
SNR	Signal-to-noise ratio	SOP	Signals-of-opportunity	
SRS	Sounding reference signal	STEC	Slant total electron content	
TDOA	Time difference of arrival	THz	Terahertz	
TLE	Two-line element	TOA	Time of arrival	
UAV	Unmanned aerial vehicle	UE	User equipment	
URA	Uniform rectangular array	UTM	Universal transverse Mercator	

characteristics of GAS anchors and targets, followed by the corresponding localization scenarios.

A. GNSS Limitations

Owing to the precise atomic clocks, GNSS was developed and first commercialized in the 1990s as a time-based localization system. The past decades have witnessed a sharp rise in GNSS-based applications and particularly GPS-based ones [15]. Nowadays, a GPS receiver is a common element in almost all mobile devices, offering navigation capability around the globe. Moreover, a standalone GPS-based localization system can provide accuracy in the range of tens of meters under ideal channel conditions. Augmenting GPS systems with reference signals, such as real-time kinematic, can bring the localization accuracy to centimeter-level [27]. Despite the high accuracy, GPS systems suffer from some critical shortcomings that can compromise the main objective of wireless applications. In outdoor rich-scattering environ-

ments and urban canyons, GPS receivers experience degraded performance, failing to provide a position fix. In the following, the main GPS drawbacks are highlighted.

- Vulnerable to jamming: Each GPS satellite broadcasts a radio signal containing its current location and the transmission time information. Under normal conditions, a GPS receiver can process the signals from multiple satellites to obtain a fix of its position. However, the reception of GPS signals can be severely degraded by either unintentional interference or intentional jamming signals. The received power of GPS signals is typically around -130 dBm on the Earth's surface, which is rather low against potential high power interference signals [28]. This will result in an intermittent reception at the GPS receiver, failing to provide a position estimate stably.
- Prone to spoofing: Spoofing is an intelligent form of malicious attacks in which the attacker broadcasts false GPS positions or fakes the GPS signal itself [29]. For instance,

false GPS positions sent from a fake automatic dependent surveillance-broadcast (ADS-B) transmitter can force an aircraft to deviate from its planned trajectory. Furthermore, an attacker can send fake GPS signals, resulting in wrong position estimates for nearby GPS receivers. In order to detect spoofing attacks and verify the real position of the target, complementary localization methods such as time-based [30] and received signal strength (RSS)-based methods [31] are employed.

Extra power and size: The size of GPS receivers, including the antenna, has been reduced over time. Nevertheless, for compact low-power devices such as IoT nodes in low-power wide-area networks [32], adding a GPS receiver will take a significant part of the overall size. This is due to the fact that GPS works on a different RF band (e.g., 1575.42 MHz) compared to the IoT band (i.e., 868 MHz). This means that an extra antenna is always needed for the GPS receiver. For example, the Telecom Design IoT modem with GPS is almost double the size of the one without GPS [33]. In addition to the extra size, GPS receivers are known to be power-hungry with respect to battery-powered IoT nodes [32], considerably reducing their lifetime. To obtain a position fix, a GPS receiver could take a few seconds of continuous processing, consuming a relatively large amount of power. Therefore, due to power and size limitations, many IoT deployments avoid including a GPS receiver in every node [5,34].

B. Characteristics of GAS Anchors

The anchors in GAS networks can be broadly classified into terrestrial anchors, which include cellular BSs and access points (APs), and non-terrestrial anchors, which include aerial anchors and space anchors, as visualized in Fig. 1. Anchor nodes are the reference nodes in the localization system. Anchors can be either used to cooperatively localize a target (e.g., uplink localization), or a target can use them as references to localize itself (e.g., downlink localization). To this end, anchors need to have a precise known location. With the advances in the multiple-input multiple-output (MIMO) technology and with aerial and space anchors entering the picture, some substantial novel characteristics can be spotted in the GAS localization system design.

• Ground Anchors: The majority of ground anchors are fixated on the ground with a precisely known location as in the case of cellular network's BSs and APs [17]. There exist cases where ground mobile robots are used as anchors; however, these cases are mainly indoor [35]. Modern ground anchors in 5G networks are designed to have a multitude of antennas, boosting AOA and AOD accuracy. Massive MIMO is one of the key technologies used in current 5G networks, in which terrestrial BSs are equipped with massive antenna arrays. The primary role of these massive antenna arrays is to increase spectral efficiency by providing high beamforming gain and spatial multiplexing for users. From the localization system perspective, such massive antenna arrays play a pivotal role in improving angle-of-arrival (AOA) and angle-of-departure (AOD) measurements by significantly elevating

- angle estimation accuracy. Ground anchors can work as transmitters and receivers. When working as receivers, they shift the processing cost from the target to the anchors or network infrastructure.
- Aerial Anchors: While terrestrial anchors tend to have fixed positions, aerial anchors have the flexibility needed to constantly adapt their positions, enabling dynamic anchor placement to achieve better localization performance. Aerial anchors can be deployed at altitudes spanning a few tens of meters up to 50 km as detailed in Fig. 1. Aerial anchors are represented by LAPs and HAPs. LAPs, a.k.a UAVs, cover altitudes from a few meters up to 5 km. HAPs, on the other hand, are quasi-stationary aerial platforms located at heights of 20–50 km above the Earth's surface in the stratospheric region of the atmosphere [7]. UAVs as aerial anchors can exploit their altitude to obtain higher line-of-sight (LOS) probability when compared to their ground counterparts. This results in significant improvements for the localization techniques, in which non-line-of-sight (NLOS) links are the primary source of error. Moreover, UAV's flexibility can be further exploited to minimize the number of anchors needed to realize a given localization accuracy [36]. Using a mobile UAV, one can use the measurements collected from multiple waypoints² to localize quasi-stationary targets [37]. In principle, these waypoints represent virtual anchor points. LAP anchors can work as transmitters and receivers, whereas HAP anchors are most commonly used as transmitters.
- Space Anchors: The deployment of space anchors takes place at LEO, MEO, or GEO orbits. However, LEO satellites, particularly, are attracting considerable research focus due to their reduced latency, broad coverage, and the possibility of deploying mega-constellations with multiple wireless technologies [38]. As a key component of NTNs, LEOs are not only expected to provide communication service thanks to the suitable altitudes (e.g., 550 km for Starlink) but also to act as a space anchor to localize users. Anchors in LEO have the advantage of covering a large part of the Earth. By deploying a constellation of a couple of hundred satellites in LEO, global coverage can be reached for a localization system [15]. However, the received power of such systems is typically a lot lower in comparison with terrestrial networks, making satellite signals mainly suitable for LOS communications. Space anchors are generally used only as transmitters; hence, corresponding users need to localize themselves.

C. Characteristics of Ground and Aerial Targets

Given that satellites are launched to a specific orbit, and they are constantly tracked [39], the targets considered in this paper, as visualized in Fig. 1, are the ones at ground and aerial segments, namely, terrestrial targets and aerial targets, respectively. Target nodes are typically power and size-limited when compared to anchors, e.g., mobile UE, armature drone, or IoT nodes [5,40]. Target localization requirements are usually set as constraints on performance metrics (cf. Subsection III-E),

²Stopping points during the UAV flying mission, where it hovers at each of them for some time before moving to the next one.

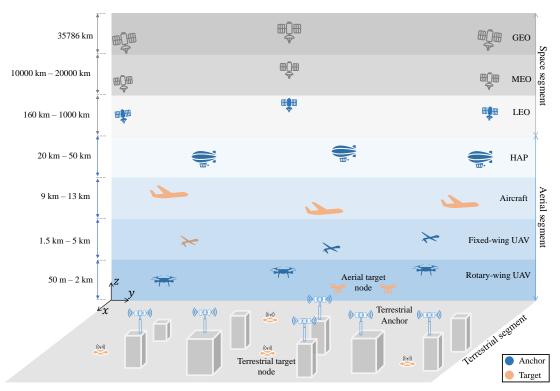


Fig. 1: A representation of the different segments and elements of integrated GAS networks and the typical altitude of the various non-terrestrial elements.

TABLE IV: Elements of the localization scenarios in GAS networks.

	Ground Anchor	Aerial Anchor	Space Anchor
Target	G2G scenario, Subsection IV-B	A2G scenario, Subsection V-B	S2G scenario, Subsection VI-B
Ground 1	Target: Connected UE, Mobile devices/vehicles Anchor: Base stations, gNB	Target: IoT nodes, Mobile devices Anchor: LAPs, HAPs	Target: IoT nodes, Mobile devices Anchor: LEO Satellites
Target	G2A scenario, Subsection IV-C	A2A scenario, Subsection V-C	S2A scenario, Subsection VI-C
Aerial	Target: LAPs, Amateur drones Anchor: Base stations, gNB	Target: LAPs, Amateur drones Anchor: LAPs, HAPs	Target: LAPs, Amateur drones Anchor: LEO Satellites

which may include accuracy, coverage, latency, complexity, power efficiency, stability, scalability, and update frequency.

• Terrestrial Targets: The main types of outdoor ground targets include UEs in cellular networks [17], connected autonomous vehicles [41], and IoT nodes [5]. The localization requirements of ground targets vary depending on the application at hand. These applications may range from autonomous vehicles and search and rescue of connected UEs, which have stringent localization requirements, to surveying, automated agriculture, and automated train operation, which require relatively relaxed localization requirements. For instance, autonomous vehicle applications require cm-level of positioning accuracy and sub-0.5 degrees for 95% of the time on average [42], whereas in some IoT applications such as package tracking, an accuracy of 100 meters is sufficient [5]. Moreover, depending on the application, ground targets might require positioning information only, like surveying and automated train operation applications, while others

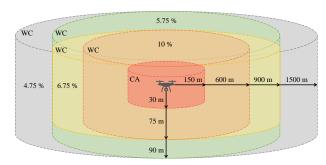


Fig. 2: The levels of drones' well-clear (WC) zones each represented by a cylinder. The red cylinder represents the collision avoidance (CA) zone. For a given drone, the percentage values showed at its WC cylinders represent the probability of another drone present in a given cylinder violating/entering the next smaller CA cylinder [3]. E.g., the probability of a drone present in the yellow CV cylinder entering the orange CV cylinder is 6.75%.

require both positioning and orientation information, like autonomous vehicles and UEs using directional beamform-

- ing in mmWave [40].
- Aerial Targets: Main examples of aerial targets are rotarywing or fix-wing drones, which are regarded as LAPs, which generally have high dynamics and relatively low altitudes (lower than 5 km). Localizing aircraft at higher altitudes such as HAPs can be neglected because of their quasi-stationary characteristics. In terms of RF connectivity, LAPs can have links with ground BSs for data exchange, air traffic management (ATM), and command and control (C&C), and hence, might act as an aerial connected UE. In addition, they could also have side links among each other in the form of device-to-device or broadcast messages. Localizing aerial targets can ensure security within a nofly zone in cases involving non-cooperative aerial targets or serve ATM in cases where GPS is not available. Noncooperative aerial vehicles include targets that do not share their GPS locations, such as amateur drones, which may raise privacy and security concerns by penetrating a no-fly zone, e.g., close to airports [43]. In non-cooperative target localization, the closer the target is to the no-fly zone, the stricter the localization requirements. In applications related to ATM, localizing requirements depend on the distancebased guidelines for UAV systems [44]. An overview of the different ATM zones is presented in Fig. 2, which can be categorized into well-clear and collision avoidance [3]. These zones can be applied to UAVs flying freely or those flying with a given route, i.e., corridors [45], aiding UAV ATM.

D. Localization Scenarios in GAS

Considering the three anchor types (ground, aerial, and space) and the two target types (ground and aerial), six positioning scenarios can be envisioned. These scenarios are ground-to-ground (G2G), ground-to-air (G2A), air-to-ground (A2G), air-to-air (A2A), space-to-ground (S2G), and space-to-air (S2A). These scenarios, along with their corresponding elements, are summarized in Table IV. While terrestrial anchors and terrestrial targets are the typical elements considered in the localization systems literature, aerial and space, particularly at the LEO orbit, introduce fundamental novelties to the conventional localization problem. For instance, a UAV can act as a supporting base station or as a user [7]. Moreover, using thousands of LEO satellites at very low LEO, below 300 km, in what is known as a mega-constellation increases the geometrical diversity of the localization system [15].

III. LOCALIZATION FUNDAMENTALS

In this section, we detail the key relevant localization fundamentals and position them in the context of GAS networks.

A. System Geometry in 3D

In GAS localization systems, it is crucial to characterize anchors and targets in a 3D space. In addition to their 3D positions, their orientation angles and velocity become highly relevant to capture their relative motion, e.g., in applications such as aerial and terrestrial autonomous vehicles [22,46].

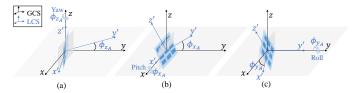


Fig. 3: An example of (a) yaw, (b) pitch, and (c) roll rotation angles performed, with right-hand rule in mind, on a uniform rectangular array of antennas. The figure shows a yaw, pitch, and roll in their corresponding positive rotation direction.

1) Reference Coordinate System: The position of a given target/anchor can be defined using the global coordinate system (GCS), whereas its orientation can be described using its local coordinate system (LCS) along with the GCS [11,40]. A widely adopted GCS is the universal transverse Mercator (UTM) coordinate system, which originates at the surface of the earth [47]. The x and y axes of the UTM system point toward the east and north of the earth, respectively, whereas the z axis is perpendicular to the earth's surface, pointing upwards, and originates at sea level. The LCS, on the other hand, has an origin centered at the antenna (array) of the corresponding anchor/target, as shown in Fig. 3. The x and z axes of the LCS point toward the anchor/target array's own horizontal and vertical directions, respectively. Subsequently, the y-axis of the LCS is perpendicular to the xz-plane, following the righthand rule, pointing towards the boresight of the array. Usually, we are interested in localizing the target in the GCS. Yet, measurements are practically taken in the LCS with respect to the anchor or the target. Hence, extra processing is usually needed to transform the LCS measurements to the GCS. The transformation from one reference frame to another is done through the translation and rotation processes [48].

The translation between the two reference frames is computed by measuring the displacement between their origins. On the other hand, the rotation between the two frames can be fully captured by the attitude angles of the LCS with respect to the GCS, known as the pitch (ϕ_x) , roll (ϕ_y) , and yaw (ϕ_7) angles [48]. To demonstrate these angles, consider the uniform rectangular array (URA) shown in Fig. 3, the pitch angle describes the vertical tilt between the y-axis of the LCS and the xy-plane of the GCS. This means that a URA with a yaxis parallel to the GCS's xy-plane is said to have a zero pitch angle. Likewise, if the antenna array is either tilted upwards or downwards, then it is said to have a positive or negative pitch angle, respectively. The roll angle, on the other hand, describes the vertical tilt between the LCS's x-axis and the xy-plane. Hence, the counterclockwise rotation of the antenna array along its y-axis corresponds to a positive roll angle and vice versa. Finally, the yaw angle describes the horizontal tilt between the LCS's y-axis and the GCS's y-axis (i.e., the north), which means that directing the antenna array westward will result in a positive yaw angle, and vice versa. The rotation between the GCS and the LCS can be represented by a matrix multiplication of three rotation matrices that correspond to the individual attitude angles. These three orientation matrices represent the pitch around the x-axis, the roll around the yaxis, and the yaw around the z-axis, as illustrated in Fig. 4. It is

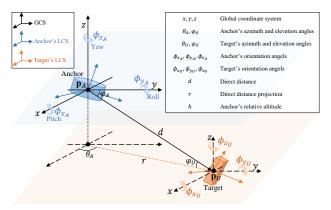


Fig. 4: A representation of the localization measurables, which include distances and angles.

worth noting that the order of applying these rotations matters and that the standard order that is conventionally followed is applying the roll first, then the pitch, and finally the yaw [48]. The rotation matrices are expressed as [49]

$$\mathbf{R}_{x_j}(\phi_{x_j}) = \begin{bmatrix} 1 & 0 & 0 \\ 0 & \cos \phi_{x_j} & -\sin \phi_{x_j} \\ 0 & \sin \phi_{x_i} & \cos \phi_{x_i} \end{bmatrix},$$
(1)

$$\mathbf{R}_{y_j}(\phi_{y_j}) = \begin{bmatrix} \cos \phi_{y_j} & 0 & \sin \phi_{y_j} \\ 0 & 1 & 0 \\ -\sin \phi_{y_j} & 0 & \cos \phi_{y_j} \end{bmatrix}, \tag{2}$$

$$\mathbf{R}_{z_j}(\phi_{z_j}) = \begin{bmatrix} \cos \phi_{z_j} & -\sin \phi_{z_j} & 0\\ \sin \phi_{z_j} & \cos \phi_{z_j} & 0\\ 0 & 0 & 1 \end{bmatrix},$$
(3)

$$\mathbf{R}_{j} = \mathbf{R}_{z_{j}}(\phi_{z_{j}})\mathbf{R}_{x_{j}}(\phi_{x_{j}})\mathbf{R}_{y_{j}}(\phi_{y_{j}}), \qquad (4)$$

where $j \in \{A, U\}$ indicates either anchor's or target's rotation, \mathbf{R}_{x_j} , \mathbf{R}_{y_j} , and \mathbf{R}_{z_j} denote the pitch, roll, and yaw rotation matrices, respectively, and \mathbf{R}_j denotes the collective rotation matrix.

2) Anchor-Target Distance and Angle Geometry: Throughout this paper, we use $\mathbf{p}_A \in \mathbb{R}^3$, $v_A \in \mathbb{R}^3$, and $\phi_A \in \mathbb{R}^3$ to denote the anchor's true position, velocity, and orientation, respectively. For instance, considering a Cartesian GCS, we have $\mathbf{p}_A = (x_A, y_A, z_A)$, $v_A = (v_{x_A}, v_{y_A}, v_{z_A})$, and $\phi_A = (\phi_{x_A}, \phi_{y_A}, \phi_{z_A})$. Here, as depicted in Fig. 4, ϕ_{x_A} , ϕ_{y_A} , and ϕ_{z_A} , respectively, correspond to the pitch, roll, and yaw attitude angles of the anchor with respect to the GCS right-hand-based frame's x, y, and z axes. Furthermore, we use $\mathbf{p}_U \in \mathbb{R}^3$, $v_U \in \mathbb{R}^3$, and $\phi_U \in \mathbb{R}^3$ to denote the user's true position, velocity, and orientation, respectively, e.g., in a Cartesian coordinate system $\mathbf{p}_U = (x_U, y_U, z_U)$, $v_U = (v_{x_U}, v_{y_U}, v_{z_U})$, and $\phi_U = (\phi_{x_U}, \phi_{y_U}, \phi_{z_U})$. A target location can be defined in the GCS with respect to an anchor position via the distance and angles between them, irrespective of their orientation, as shown in Fig. 4. The direct distance d, in 3D, between

a target located at $\mathbf{p}_U = (x_U, y_U, z_U)$ and an anchor at $\mathbf{p}_A = (x_A, y_A, z_A)$ can be written as

$$d = \|\mathbf{p}_A - \mathbf{p}_U\|$$

$$= \sqrt{(x_A - x_U)^2 + (y_A - y_U)^2 + (z_A - z_U)^2}.$$
 (5)

The projection of d on the xy-plane of the GCS provides the 2D distance $r = (z_A - z_U)/d$, as shown in Fig. 4. The GCS 3D angles, represented by the horizontal azimuth angle θ_A and the vertical elevation angle φ_A with respect to the anchor in the GCS, can be geometrically expressed as follows

$$\begin{bmatrix} \theta_A \\ \varphi_A \end{bmatrix} = \begin{bmatrix} \tan^{-1} \left(\frac{y_U - y_A}{x_U - x_A} \right) \\ \sin^{-1} \left(\frac{z_U - z_A}{d} \right) \end{bmatrix}, \tag{6}$$

where $\theta_A \in (-\pi, \pi)$ is the horizontal counterclockwise angle from the *x*-axis, and $\varphi_A \in (-\pi/2, \pi/2)$ is the vertical counterclockwise angle from the *xy*-plane. From a target perspective, these geometric azimuth and elevation angles would thus be

$$\theta_U = \begin{cases} \theta_A - \pi, & \theta_A > 0 \\ \theta_A + \pi, & \theta_A \le 0, \end{cases}$$
 (7)

$$\varphi_U = -\varphi_A \,, \tag{8}$$

in the case of LOS operation. While (6) presents the GCS geometric angles, irrespective of the orientation, the geometric azimuth and elevation angle measurements, in practice, will be affected by the orientation of the measuring antenna array at the anchor/target since they are measured in the local coordinate system. Hence, we need to estimate or know the orientation of the measuring device to extract the GCS azimuth and elevation angles from the LCS AOA/AOD measurements, cf. Section III-B4. Here, we will consider an example where we are interested in extracting the GCS position of the UE from a range and an AOA measurement that is measured at the anchor. The same method can be used for AOD measurements at the anchor and for AOA/AOD measurements at the user, assuming that the orientation of the user is known. The anchor AOA measurement can be theoretically computed as

$$\begin{bmatrix} \theta_A' \\ \varphi_A' \end{bmatrix} = \begin{bmatrix} \tan^{-1} \left(\frac{[\mathbf{R}_A^{\top} (\mathbf{p}_U - \mathbf{p}_A)]_2}{[\mathbf{R}_A^{\top} (\mathbf{p}_U - \mathbf{p}_A)]_1} \right) \\ \sin^{-1} \left(\frac{[\mathbf{R}_A^{\top} (\mathbf{p}_U - \mathbf{p}_A)]_3}{d} \right) \end{bmatrix}, \tag{9}$$

where $[\mathbf{a}]_i$ indicates the *i*-th element of vector \mathbf{a} . In (9), θ_A' and φ_A' correspond to the array's relative counterclockwise horizontal and vertical AOAs with respect to the array's *x*-axis and *xy*-plane, respectively. Hence, the user's position can be theoretically computed from anchor's LOS range, *d*, and AOA measurements, θ_A' and φ_A' , as

$$\mathbf{p}_{U} = \mathbf{p}_{A} + d \mathbf{R}_{A} \begin{bmatrix} \cos \theta'_{A} \cos \varphi'_{A} \\ \sin \theta'_{A} \cos \varphi'_{A} \\ \sin \varphi'_{A} \end{bmatrix}. \tag{10}$$

B. Localization Measurables

Radio localization is fundamentally based on the spatial information implicitly conveyed by radio waves as they propagate through the wireless channel. This spatial information is impacted by the stochastic nature of the wireless channel as well as the transceiver noise and hardware imperfections. These effects can be observed via power, time, and angle measurements [50]. In the following, we first define measurement resolution, resolvability, and accuracy. Subsequently, we detail the localization measurables.

- 1) Measurement Resolution, Resolvability, and Accuracy: Before delving into the individual measurables, it is worth noting that in model-driven localization methods, a great emphasis is placed on the accuracy, resolution, and resolvability of the acquired measurables. In this paper, we use the following definitions.
- Resolution is the smallest distinguishable change in a quantity, e.g., time, power, or angle, that can be detected by a measuring instrument, such as a transmitter or receiver. Hence, it describes the degree of detail a measurement can have. Resolution is mainly affected by the receiver's hardware, such as the analog-to-digital converter (ADC)'s quantization error [51].
- Resolvability of components in a measurement is the ability to distinguish/resolve the various multipath components, i.e., LOS and NLOS components, that constitute the measured quantity. Resolvability is bounded by the available bandwidth for time-based measurements, the coherent integration time for frequency/Doppler measurements, and the size of the antenna array for angle-based measurements³. If the measurements have low resolvability in the corresponding time, Doppler, or angle domain, then various path components, e.g., LOS and NLOS, will be amalgamated, limiting the localization performance, irrespective of the signal-to-noise ratio (SNR), unless super-resolution methods are applied [53].
- Accuracy, by definition, is a measure of how close the acquired measurement is to the true value of the measurable. The accuracy of measurements is affected by many factors, including both measurement resolution and resolvability, as well as the number of antennas used, the bandwidth, and the SNR. To assess the quality of measurements, one needs to consider both accuracy (bias) and precision (variance), which are often combined in the form of root-mean-square error (RMSE) [11,16].
- 2) Power Measurements: Localization using power measurements relies on RSS, which is a measure of the received radio signal power. Measuring RSS is a natively supported feature in modern wireless transceivers. Thanks to their intrinsic simplicity, RSS-based methods have been widely adopted in the literature, offering cost and energy-efficient localization solutions [54]. RSS-based localization can be roughly classified into range-free and range-based methods [54]. Range-free methods do not include distance measurements but rather rely

on connectivity or signal propagation aspects between anchors and target nodes [55]. Range-based methods, on the other hand, use RSS measurements to estimate the direct distance between an anchor and a target. The time-averaged received power, $P_{\rm Rx}$, at an anchor located at a direct distance d from a transmitting target node, can be generally expressed, in dBm, as [54,56]

$$P_{Rx} = P_{Tx} + \underbrace{C - 10n_p \log_{10} \left(\frac{d}{d_0}\right)}_{\text{Path gain}, G_p} + \mathcal{X}, \ \forall d \ge d_0, \tag{11}$$

where P_{Tx} is the transmit power, in dBm, n_p is the pathloss exponent, which reflects the way the received power decays with distance, and d_0 is the reference distance after which the far-field lies, and consequently, the received power model in (11) is valid. In (11), X is a zero-mean normal random variable with a standard deviation σ_X (in dB), representing the shadowing effect. The pathloss exponent, n_p , and shadowing standard deviation, σ_X , depend on the considered scenario from Table IV (e.g., cf. Section V-A2). The term C, in (11) is a distance-independent term that collects constant factors between a given transmitter and a receiver such as the carrier frequency and the transmitter and receiver antenna gains, respectively denoted by G_{Tx} and G_{Rx} . The gain of the antenna differs with the type of antenna (antenna pattern), which is particularly formulated as⁴

$$G_j(\boldsymbol{\psi}_j') = \eta \cdot \frac{4\pi A_e(\boldsymbol{\psi}_j')}{\lambda^2}, \ j \in \{\text{Tx}, \text{Rx}\},$$
 (12)

where $\psi'_j = [\theta'_j, \varphi'_j]$ comprises the array's relative horizontal (azimuth) and vertical (elevation) angles, λ is the wavelength of the carrier frequency, η is the antenna efficiency, and A_e represents the effective aperture of the antenna.

In practice, receivers express the RSS as a scaled version of P_{Rx} , i.e., $RSS = \mathcal{R}(P_{Rx})$, where $\mathcal{R}(.)$ is the received power to RSS transduction function, which ideally provides a rescaled version of the received power [54]. In order to overcome any hardware impairments, a calibration process is needed to accurately rectify the power-to-RSS transduction function $\mathcal{R}(.)$ [58]. Assuming a calibrated receiver, the RSS-based estimated direct distance between an anchor and a target is

$$\hat{d} = d_0 10^{\frac{P_{\text{Tx}} + C - P_{\text{Rx}}}{10n_p}} + e_{\text{RSS}},$$
 (13)

where e_{RSS} is the RSS measurement error typically caused by the shadowing effect X and the uncertainty of antenna gains in the direction of departure and arrival. It is worth noting that (13) provides a distance estimate for a given observation of P_{Rx} . However, for a given distance d, this observation is, in fact, a random variable statistically influenced by the shadowing effect X.

³Some works in the literature, particularly in the radar community, use the term measurement resolution to describe the resolvability/separability of radio paths in the time, Doppler, or angle domain [52].

 $^{^4}$ The dependency on the antennas' gains implies that C depends on the transmitter's antenna orientation relative to the receiver's antenna orientation [57]. The effect of the relative antenna orientation needs to be measured and calibrated.

3) Time Measurements: The core idea of a radio time measurement lies in the transmitter's and receiver's ability to define the exact timestamp of a radio signal's transmission and reception, respectively. Defining transmission timestamp at a transmitter requires a stable local clock with bounded drift, i.e., an oscillator that can provide an accurate and stable clock signal [59], whereas defining the reception timestamp at the receiver, commonly known as time of arrival (TOA), requires not only a stable clock at the receiver but also an accurate TOA estimation technique. Examples of TOA estimation techniques are peak detection techniques and statistical goodness-of-fitbased techniques [60]. The received SNR and signal's bandwidth are key design parameters for better TOA estimation. Assume t_0 is the transmission time at the transmitter, and t_r is the true TOA at the receiver. The estimated direct distance between the transmitter and the receiver can be expressed as

$$\hat{d} = c \,\Delta t + e_{\text{TOA}} \,, \tag{14}$$

where c is the speed of light, $\Delta t = t_r - t_0$ is the true propagation time, and e_{TOA} is the time measurement error resulting from multipath, e.g., positive bias due to the NLOS propagation, as well as the clock offset between transmitter's and receiver's clocks. Synchronizing a transmitter and a receiver, e.g., an anchor and a target node, to realize accurate TOA measurements is not an easy task due to clock imperfections, which result in a drift that depends on environmental factors such as temperature and pressure [59]. One way to avoid the need to synchronize target nodes, e.g., non-cooperative targets, with anchors is time differences of arrival (TDOA), which relies on synchronized anchors to measure the difference in distance using TOA at several anchors. The corresponding difference in distance, d_{12} , from a target to two anchors can be written as [50]

$$\hat{d}_{12} = (t_{A_1} - t_{A_2})c + e_{\text{TDOA}}, \tag{15}$$

where t_{A_1} and t_{A_2} denote the TOA at anchor A_1 and anchor A_2 , respectively, and e_{TDOA} is the TDOA measurement error, which similar to e_{TOA} , is influenced by the multipath propagation, and the clock offset between the anchor's clocks. Another widely adopted time-based ranging method is round-trip time (RTT) [17]. Unlike TDOA and TOA, RTT counteracted the synchronization requirements by performing back-and-forth time measurements per individual anchor. The main challenge with round-trip time is determining a good estimate of the processing duration at the receiver, which is hardware-dependent, making it only suitable for cooperative targets.

4) Angle Measurements: Antenna arrays can be exploited to measure the AOD and the AOA between an anchor and a target node. In order to localize a target in 3D space using angle measurements, one needs to estimate both azimuth and elevation angles, as discussed in Section III-A and as shown in Fig. 4. To estimate both angles at a single anchor, it should be equipped with a 2D array, e.g., URA. This means that in order to measure both the azimuth and elevation angles of the uplink-AOA (UL-AOA) or the downlink-AOD (DL-AOD), the anchor node should be equipped with a 2D array. Likewise, if the DL-AOA or the UL-AOD are to be measured, then the user should be equipped with a 2D array. It is worth remembering

that for these angles to be useful for positioning purposes, the attitude angles (i.e., pitch, roll, and yaw) of the measuring node should be known. In cases where the attitude angles of the measuring device are not known, then co-estimation of both the orientation and the position of the target should be conducted in what is commonly known as 6D localization [61]. In these cases, extra information offered by other anchors can be utilized to solve the estimation problem.

The AOA is estimated by measuring the phase differentials across the array, as signals arrive at each antenna element with a slight delay compared to the array's phase center. These minor delays result in phase rotations, which can be measured when an RF chain is available at each antenna element [62]. While there can be more than one propagation path, only the LOS path is usually of interest for angle-based localization. This LOS path can be extracted by, e.g., the MUSIC algorithm [63] or ESPRIT algorithm [64]. The AOD is often estimated by sending different directional beams in different directions/AODs (e.g., with different azimuth and elevation angles when the transmitter is equipped with a 2D array). At the receiver's side, RSS measurements are conducted for the various beams. The AOD is then estimated corresponding to the beam with maximum strength [65]. In case the receiver has access to the complex beam responses and beam weights, more sophisticated methods can be applied, such as beamspace ESPRIT [66] or orthogonal matching pursuit [67].

5) Doppler Measurements: In the case of a mobile localization system, the estimation of the 3D velocity of UEs is paramount for many applications, e.g., autonomous vehicles. To estimate the velocity of a mobile UE, we usually rely on Doppler measurements, which measure the frequency shifts in the carrier frequency, and sub-carrier frequencies in OFDM-based systems, from the nominal frequency of operation (i.e., the original carrier frequency) due to the relative radial velocity between the UE and the anchor node [68]. The relative radial velocity⁵ between a given anchor and a UE can be written as

$$\Delta v = (\mathbf{v}_A - \mathbf{v}_U)^{\top} \frac{\mathbf{p}_A - \mathbf{p}_U}{\|\mathbf{p}_A - \mathbf{p}_U\|}.$$
 (16)

In a LOS scenario, the estimated frequency of the received signal, \hat{f}_r can be formulated as

$$\hat{f}_r = f_c \left(1 + \frac{\Delta v}{c} \right) + e_{Dr} \,, \tag{17}$$

where f_c is the carrier frequency and e_{Dr} is the Doppler measurement error which is mainly caused by the carrier frequency offset (CFO) between the transmitter and the receiver. Hence, in order to measure the contribution of the radial velocity to the frequency shift, we need to estimate and compensate the CFO. In NLOS scenarios, where the signal bounces off multiple objects before reaching the receiver, (17) must be altered to include the relative radial velocities between the various objects in the environment. Note that the Doppler measurements do not only possess information about the relative velocity of the UE, but also its position. Hence,

⁵The relative radial velocity is also commonly referred to as the LOS velocity [68].

the availability of Doppler measurements can lead to enhanced positioning solutions [68,69].

6) Joint measurables: Methods exist that perform joint estimation in several domains, e.g., TOA and AOA [70]–[72], which provides improved accuracy at the cost of increased processing complexity. Compared to using a single measurement domain, multi-domain joint time and angle estimation methods can fully exploit both temporal and spatial diversity in resolving multipath components [73]. In order to jointly estimate time and angle information in multipath channels, the selection of bandwidth and array size can be jointly optimized to ensure time and angle resolvability. The joint estimation of angle and time can be done using sub-space-based methods or maximum-likelihood-based methods [74]. The former methods tend to be sub-optimal and require relatively high SNR. In contrast, the latter methods usually impose major computational complexity as they need to estimate time and angle iteratively.

C. Snapshot Localization Methods

Once ranges, range differences, Doppler, and/or angles are measured as discussed in Section III-B, a data fusion step follows, in which measurements from/at different anchors are combined to obtain an estimate of the target node location. Broadly speaking, localization methods can exploit localization measurables in a data-driven or model-driven manner [52]. In data-driven methods, such as RFs fingerprinting, machine learning is employed to take advantage of the wireless channel's rich features to map received signal measurements spatially. In this context, no closed-form equation or easyto-model mathematical framework is used. The model-driven methods, on the other hand, harness stochastic models describing the relation between the received radio signal and the signal propagation geometry from/to the target location. In general, the accuracy of model-driven methods depends on the ability to extract the LOS path between an anchor and a target, making the multipath nature of the wireless channel one of the major localization error sources for LOS-based methods [50,52].

The sensor fusion architectures in model-driven methods can be generally categorized into centralized (also known as tightly-coupled) or decentralized (also known as looselycoupled) integration schemes. Centralized integration methods usually rely on a single central filter that fuses the positioning measurables from various anchors. In contrast, decentralized fusion algorithms fuse the individual/independent positioning estimates of the participating anchors into a single positioning estimate. Hence, by design, decentralized integration can only work for technologies that can estimate the position using a single anchor node, e.g., hybrid range and angle. A common practice in all measurement fusion methods is first to form a system of equations, e.g., in matrix form, linking the unknown target position to the measurements and anchors positions. Subsequently, the resulting set of system of equations can be solved to estimate the unknown target position. Broadly speaking, the system of equations' solution can be obtained either using a statistical approach based on the maximum likelihood estimator or using an algebraic approach based on the least-squares estimator [50,75]. In the following, we detail the different model-driven data fusion methods used to construct the system of equations.

- Ranging-based multilateration: Ranging-based multilateration (MLAT) uses the estimated distances between a target node and at least four different anchors in order to perform 3D localization, as illustrated in Fig. 5a. Once a sufficient number of distance estimates between anchors and the target are collected, one can formulate a system of equations representing the MLAT estimation problem [50]. Subsequently, the target's position is determined by solving the MLAT estimation problem using (non)linear least squares estimators or maximum likelihood estimators. MLAT is usually utilized when only range measurements are available, i.e., no angle-based measurements. Also, MLAT is preferred even if angle measurements are available in scenarios where the distance between the anchor node and the UE is relatively large, as in GNSS, where small angle measurement errors cause high positioning errors. It is worth noting that the relationship between range measurements and position states is non-linear, and hence, linearization errors are incurred when utilizing linear filters. These linearization errors are magnified when the UE is close to the anchor node and can be neglected for long UE-anchor distance [76].
- Range-difference-based multilateration: In range-differencebased MLAT, range-difference between at least four anchors is needed to estimate a 3D target's position, as shown in Fig. 5b. Similar to ranging-based MLAT, a system of equations can be formulated using range difference measurements, e.g., using TDOA [75]. Range-difference-based MLAT can be used in the same scenarios where MLAT can be used, but requires tight synchronization between the anchor nodes. This method is more popular in cellular positioning, where BSs are expected to be communicating and synchronized with each other. Like ranging-based MLAT, this method incurs linearization errors when linear filters are utilized. The linearization error increases when the UE is close to either of the anchor nodes. Hence, the lowest linearization error occurs when the UE is right in the middle between the anchor nodes [77].
- *Triangulation*: The basic principle of triangulation, illustrated in Fig. 5c, is to measure the angle between the target and at least two anchors to perform 3D localization. In triangulation algorithms, we construct a set of linear equations as a function of the target position, (x_U, y_U, z_U) , where each anchor may contribute to two equations of the equations set [78]. Triangulation is usually utilized when only angle measurements are available, e.g., no rangebased measurements. Similar to its MLAT counterparts, this method incurs high linearization errors when used with linear filters [79].
- Hybrid distance-angle: The main advantage of hybrid fusion methods is their ability to provide a target location estimate using a single anchor [50], as shown in Fig. 5d. For instance, an anchor with a URA can perform AOA estimates and combine them with ranging-based measurements, e.g., TOA, to obtain a target's position. Additionally, hybrid positioning

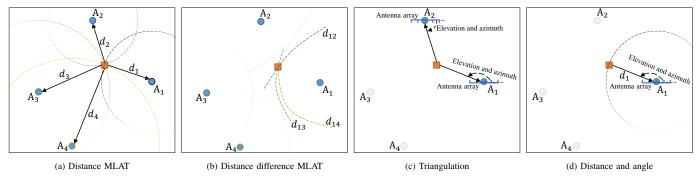


Fig. 5: Primary snapshot localization methods.

is preferred when using decentralized linear integration schemes, as it minimizes the linearization errors caused by their MLAT and triangulation counterparts [80].

D. From Snapshot to Tracking

Snapshot localization is realized when the state of interest x, i.e., the position of the target, is assumed to be stationary and not in motion. Tackling an estimation problem with dynamic states is known as a tracking problem. Discrete-time tracking implementations usually use a discrete-time index denoted by k. State estimation in tracking problems comprises a recursive iteration of state prediction and correction in what is commonly known as a filter [81]. In the state prediction stage, the filter predicts the current epoch's state x_k^- through a transition model $f(x_{k-1}^+, u_k)$ that utilizes the last epoch's estimated state x_{k-1}^+ and the current epoch's control inputs, denoted by u_k . Here, u_k usually comprises the control signals that control the motion of the UE, like the steering wheel and the gas pedal signals of land vehicles or the throttle and the wing control signals of aerial vehicles. In case these signals are not available, measurements from motion sensors that are attached to the UE, like accelerometers, gyroscopes, odometers, or magnetometers, can act as the control inputs \boldsymbol{u}_k . Note that the predicted and corrected states are usually distinguished by negative and positive superscripts, respectively. The correction stage is conducted through the utilization of the measurements vector z, which might comprise range, angle, position, orientation, and/or velocity measurements. There are many estimators proposed in the tracking literature that are utilized for positioning purposes. The most famous among them is the family of Kalman filters (KFs) and particle filters [82]. One of the main differences between these filters is that the KF propagates a single belief/hypothesis of the state of the system, which comprises the mean, x, and the variance, **P**, of the state. On the other hand, the particle filter entertains multiple beliefs/hypotheses of the states of the system, i.e., the particles. Kalman filtering is a Bayesian estimator that is considered to be BLUE (best linear unbiased estimator) when the following criteria are met:

- Transition $f(\cdot)$ and measurement $h(\cdot)$ models are linear.
- The noise of the process and the measurements are independent zero-mean (unbiased) white Gaussian noises with perfectly known covariance.

If the criteria above are met, the KF is optimal in the sense that it will result in the minimum state estimation covariance compared to other linear estimators. If any of the optimality criteria are breached, then the KF is considered sub-optimal [81]. The KF has evolved over the years to tackle such scenarios of sub-optimality better. For instance, the extended KF and the unscented KF have been proposed to address the linearity restrictions. The particle filter, on the other hand, does not assume the linearity of the models nor a specific distribution of the noises [83]; hence, there is no guarantee of optimality. Instead, it relies on the propagation, weighting, and redistribution of a large number of particles, which hold different hypotheses of the state of the system [83]. It is worth noting that the propagation of a high number of beliefs will lead to higher computational complexity compared to the various KF implementations [83].

E. Key Performance Metrics

The main objective of any localization system, whether localizing an aerial or a terrestrial target, is to estimate the target's position with the minimum error possible. This objective intuitively makes localization accuracy a primary performance metric. Nonetheless, other localization KPIs play an essential role in the localization system design. In the following, we detail these localization KPIs.

• Localization Accuracy: The accuracy as a localization KPI is often presented as a metric combining both the accuracy, i.e., bias, as well as the precision, i.e., variance. Throughout this tutorial, we use the term localization accuracy to refer to both bias and variance unless otherwise mentioned. The localization error in 3D is defined as the Euclidean distance between the true position, $\mathbf{p}_U \in \mathbb{R}^3$ and its estimate, $\hat{\mathbf{p}}_U$, which can be expressed as

$$\mathcal{E}_{\mathbf{p}_U} = \|\mathbf{p}_U - \hat{\mathbf{p}}_U\|, \tag{18}$$

where $\|\cdot\|$ is the L²-norm. The equation in (18) calculates a snapshot localization error from a single observation. In order to evaluate the localization system accuracy (bias and variance) from multiple spatial or temporal observations, the RMSE is typically used, which is given by

$$\mathcal{E}_{\mathbf{p}} = \sqrt{\mathbb{E}\left[\|\mathbf{p}_{U} - \hat{\mathbf{p}}_{U}\|^{2}\right]},$$
(19)

where $\mathbb{E}[\cdot]$ is the expected value. Alternatively, error percentiles can be used to represent the performance over

multiple time/space location observations, which can be obtained using the cumulative distribution function (CDF) of the localization error. For example, some applications care about the percentiles of the error, i.e., what error can be attained for 90% or 99% of the time/space. One of the most widely adopted benchmarks for localization accuracy is the Cramér-Rao lower bound (CRLB), which not only provides a lower bound for the location estimator but can also be used as an indicator for the design and optimization of localization systems. In particular, the CRLB defines a lower bound for the variance of any (asymptotically) unbiased estimator [84].

- Localization Coverage: Since most of the localization methods involve measurements from several anchors collectively working to estimate a target's position, the localization coverage could be defined as the region where localization (obtaining a position fix) is possible or the region where the geometric dilution of precision (GDOP) is sufficiently good, or the region where the localization error is below a defined threshold. A more strict definition of the localization coverage can be obtained by setting an upper bound on the range/angle estimator error per anchor, which could be a factor of the CRLB [85].
- Localization Latency: The localization latency metric is the time duration between the trigger of the localization process and the time at which the target position is estimated. This time is mainly affected by three factors, namely, the propagation time between the transmitter and the receiver, the processing time, and the measurement aggregation time. Latency caused by propagation time is usually negligible for ground and aerial operation, as it is in the order of (sub)micro-seconds. On the other hand, S2G and S2A propagation time is usually in the order of milliseconds and thus cannot be neglected. Processing time is the time needed to process a set of measurables, which usually scales with the system's complexity and is a function of the signaling protocols deployed. Lastly, as the name suggests, aggregation time accounts for the time taken to accumulate enough poisoning measurables to estimate the position. For instance, in cases where only range measurements are available from a single UAV or LEO satellite, measurements should be accumulated over an extended period of time to compute a single anchor MLAT solution. It is worth noting that the latency KPI is mainly important for mobile target applications, in which an error caused by small localization latency scales up with the target's speed.
- Localization Power-Efficiency: Tracking power consumption is crucial to ensure an energy-efficient localization process, especially when a mobile anchor is used or in case several iterations of signaling are needed between the anchor and the corresponding target. This metric is particularly important in IoT applications, e.g., assets tracking, where prolonging the IoT node's battery life is a priority, even if it means compromising the localization accuracy by several meters [5]. The localization power efficiency as a metric is directly related to the communication computational complexity of the localization algorithm at hand, i.e., complex localization algorithms with long convergence time are power-hungry.

TABLE V: 5G radio technology.

Technology	Frequency band	Bandwidth	Range	
5G-NR FR1	sub-6 GHz	20 - 100 MHz	~ 1 km [88]	
5G-NR FR2	27-71 GHz	400 MHz	~ 190 m [88]	

- Localization Stability: As we explore high-frequency bands such as mmWave and THz that function in narrow beams, considerable attention must be paid to beam misalignment. Beam pointing errors can cause deafness, which results in localization outage or loss of tracking. The localization stability represents the rate at which beam misalignment occurs [11]—this misalignment rate increases in the case of a mobile anchor/target.
- Localization Scalability: The scalability metric measures the localization system's ability to scale with the number of targets, considering anchors' placement, processing time, and power constraints [11,86]. For instance, a localization system that uses TDOA scales better with the number of targets compared to a localization system based on RTT. This is due to the fact that RTT requires dedicated two-way communications between anchors and the targets, as opposed to TDOA, which can be done based on broadcast signals.

IV. LOCALIZATION WITH GROUND ANCHORS (5G)

In the past, localization with ground anchors has been synonymous with indoor localization due to the short-range operation of contemporary technologies such as Wi-Fi, ultrawideband, Bluetooth, and ZigBee. Such a short range of operation made outdoor deployment prohibitive in nature [87]. Outdoor localization through cellular ground anchor nodes has been investigated using 3G and 4G technologies, yet it was limited to low-accuracy positioning. This is mainly due to the fact that 3G and 4G positioning measurables were of low quality due to the low bandwidth and number of antennas utilized [17]. On the other hand, the 5G of cellular networks holds great promise to realize high-precision positioning thanks to the use of mmWaves and massive antenna arrays [17]. Hence, this section will discuss the general 5G system model and the various system design aspects, research topics, opportunities, and challenges when localizing ground and aerial UEs using terrestrial 5G anchors.

A. System Model

1) Radio Technology: 5G-NR (new radio) have access to two frequency ranges, FR1 (sub-6 GHz) and FR2 (mmWave) as presented in Table V. FR1 is designed for medium to long-range communications while FR2 (27 – 71 GHz) focuses on short-range, deep-urban environment communications [89]. Note that distance ranges in Table V assume an SNR of 0 dB [88]. FR1 radios have access to a bandwidth ranging from 20 to 100 MHz. On the other hand, thanks to mmWaves, FR2 5G terrestrial networks (TNs) have a large bandwidth that may reach up to 400 MHz, which allows for accurate time-based measurements (TOA, TDOA, and RTT) and the ability to resolve multipath signals in the time domain [90]. Moreover,

5G FR2 BSs are equipped with a large number of antennas, enabling accurate UL-AOA and DL-AOD measurements and resolving multipath signals in the angular domain [90]. It is worth noting that UE antenna arrays are expected to house significantly fewer antenna elements than BS arrays [91]. Hence, they might be able to perform DL-AOA and UL-AOD measurements but at a lower accuracy. In the case of having access to both UE-based and BS-based angle measurements, the orientation of the UE can be estimated. 5G-NR is set to achieve a maximum latency of 1 ms, which is crucial for delaysensitive applications like autonomous vehicle navigation [90]. Low latency communication is even more important when the UE is moving at high speed. For instance, if the UE is in a vehicle that is driving at 100 km/hr, then a latency of 10 ms would result in a constant positioning error of 27.7 cm, not counting other sources of error. Finally, 5G FR2 small cells are designed to be densely deployed with an inter-cell distance of 200 - 500 m [92], which gives more chances to have a LOS communication with multiple BSs simultaneously, reducing positioning errors.

In order to estimate the various 5G localization measurements, two pilot signals were specifically designed, namely the downlink positioning reference signal (PRS) and the uplink sounding reference signal (SRS). The PRS and SRS were originally introduced by the 3GPP for LTE localization [87]. 3GPP Release 16 enhanced the functionality of PRS and SRS by allocating higher bandwidth to them (i.e., 400 MHz) [93]. PRS and SRS can have four different sub-carrier comb sizes/patterns, i.e., (2, 4, 6, and 12). The comb pattern indicates the spacing between the resource elements (sub-carriers) utilized by the reference signals within a single time slot. A denser comb pattern, e.g., comb size of 2, means that the UE will have access to multiple measurements from multiple sub-carriers per orthogonal frequency division multiplexing (OFDM) symbol in a given resource block, enhancing the accuracy of the positioning measurements [65].

2) Channel Model: In order to introduce a general model for a 5G FR2 localization system, we consider a MIMO system with $N_{\rm Tx}$ transmitting antennas, $N_{\rm Rx}$ receiving antennas, $N_{\rm RF}$ RF chains, and G paths between the transmitter and the receiver. Moreover, we consider the transmission of 5G OFDM frames comprising K_s sub-carriers, and L_s symbols. The individual paths, sub-carriers, and OFDM symbols are denoted by g = 0, ..., G - 1, $k = 0, ..., K_s - 1$ and $l = 0, ..., L_s - 1$, respectively. The duration of each OFDM symbol is T_{Sym} = $T_{\rm CP} + T_p$, where $T_{\rm CP}$ and T_p are the cyclic prefix period and the actual symbol period, respectively. The OFDM subcarrier spacing is thus $\Delta f = 1/T_p$. The received signal at the receiver, denoted by $y_{k,l} \in \mathbb{C}^{N_{RF}}$ constitutes the transmitted complex pilot symbol, denoted by $s_{k,l} \in \mathbb{C}$, modulated by the precoding vector $f_l \in \mathbb{C}^{N_{\text{Tx}}}$, the channel impulse response $H_{k,l} \in \mathbb{C}^{N_{\text{Rx}} \times N_{\text{Tx}}}$, and the combining matrix $W_l \in \mathbb{C}^{N_{\text{Rx}} \times N_{\text{RF}}}$, and corrupted by the thermal noise $n_{k,l} \sim CN(0, W_l^H W_l \sigma_{p_v}^2)$ at the receiver [94]:

$$y_{k,l} = W_l^{\mathsf{H}} H_{k,l} f_l s_{k,l} + n_{k,l} . {20}$$

The precoding vector f_l and the combining matrix W_l are designed and optimized by the operator to either serve com-

munications or localization purposes. On the other hand, the channel between the transmitter and the receiver is mainly determined by the surrounding environment. It is worth noting that such a notion of complete uncontrollability over the channel is currently being challenged by the emergence of RIS, which will be discussed in Section VII. The general channel model between a ground UE and a BS for the *k*-th sub-carrier and the *l*-th OFDM symbol while aggregating *G* paths in a MIMO system is

$$\mathbf{H}_{k,l} = \sum_{g=0}^{G-1} \alpha_g \underbrace{e^{-J2\pi(f_c + k\Delta f)\tau_g}}_{\text{path delay}} \underbrace{e^{J2\pi f_c \nu_g lT_{\text{Sym}}}}_{\text{frequency shift}} \mathbf{a}_{\text{Rx}}(\boldsymbol{\psi}'_{\text{Rx}_g}) \mathbf{a}_{\text{Tx}}^{\top}(\boldsymbol{\psi}'_{\text{Tx}_g}),$$

$$\underbrace{\mathbf{a}_{\text{Rx}}(\boldsymbol{\psi}'_{\text{Rx}_g}) \mathbf{a}_{\text{Tx}}^{\top}(\boldsymbol{\psi}'_{\text{Tx}_g})}_{\text{steering vectors}},$$
(21)

where $\alpha_g \in \mathbb{R}$ is the channel attenuation factor, τ_g is the delay of the g-th path, f_c is the carrier frequency, ν_g is the frequency shift factor caused by the g-th path, \mathbf{a}_{Rx} and \mathbf{a}_{Tx} are the receiver and transmitter steering vectors, respectively, $\psi'_{Rx_g} = [\theta'_{Rx_g}, \varphi'_{Rx_g}]$ comprises the horizontal (azimuth) and vertical (elevation) angles of arrival relative to the receiver's orientation, and $\psi'_{Tx_g} = [\theta'_{Tx_g}, \varphi'_{Tx_g}]$ comprises the horizontal and vertical angles of departure relative to the transmitter's orientation. In the following, we use g = 0 to indicate the LOS path, if any, and $g \neq 0$ to indicate the NLOS components. The channel gain, α_g , is modeled as follows

$$\alpha_g = 10^{\left(\frac{G_p}{20}\right)},\tag{22}$$

where G_p is the path gain in dB from (11). The path delay is computed as $\tau_g = d_g/c + \tau_b$, where d_g is the length of the g-th path and τ_b is the time synchronization bias between the transmitter and the receiver. In the case of LOS communications, the LOS path length is the direct distance between the anchor and the target and is computed as given in (5). In case of an NLOS path g with a number of bounces $M_g \ge 2$, the distance will be

$$d_g = \|\mathbf{p}_A - \mathbf{s}_{g,1}\| + \|\mathbf{s}_{g,M_g} - \mathbf{p}_U\| + \sum_{m=1}^{M_g - 1} \|\mathbf{s}_{g,m} - \mathbf{s}_{g,m+1}\|, (23)$$

where, $s_{g,m}$ is the position of the m-th scatterer of the g-th path, \mathbf{p}_A is the position of the BS, and \mathbf{p}_U is the position of the UE as stated in Section III-A. The frequency shift component in (21) is caused by the net difference of the radial velocity between the receiver and the last incident point of the given path, i.e., the Doppler shift, and the CFO factor between the oscillators of the transmitter and the receiver [94]. The radial velocity is a measure of how the transmitter and the receiver are moving away/closer from/to each other. Hence, relative tangential motion between the two does not contribute to the Doppler shift. In the case of localization with ground anchor nodes, the BS can be assumed to be stationary. Hence, only the velocity of the UE, denoted by $\mathbf{v}_U = [v_{x_U}, v_{y_U}, v_{z_U}]^{\mathsf{T}}$, is taken into consideration. The CFO factor is, in principle, unknown and is denoted by δ_f . Based on (17), the frequency

shift factor of the *g*-th path, assuming that the reflectors are stationary, can be computed as

$$\nu_g = \frac{\mathbf{v}_U^{\mathsf{T}} \mathbf{u}_g}{c} + \delta_f, \tag{24}$$

where

$$u_g = u(\theta_g, \varphi_g)$$

$$= [\cos(\theta_g)\cos(\varphi_g), \sin(\theta_g)\cos(\varphi_g), \sin(\varphi_g)]^{\top}$$

$$= \frac{s_{g,M_g} - \mathbf{p}_U}{\|s_{g,M_g} - \mathbf{p}_U\|}.$$
(25)

Here, θ_g and φ_g are the relative horizontal and elevation angles between the receiver and s_{g,M_g} , the last incident point of the g-th path, in the GCS.

The steering vectors $\mathbf{a}_{Rx}(\boldsymbol{\psi}'_{Rx_g})$ and $\mathbf{a}_{Tx}(\boldsymbol{\psi}'_{Tx_g})$ can be formulated by assuming a local coordinate system, centered around the center of the given array, as discussed in Section III-A. Hence, the various array elements will transmit/receive the signal with a positive or a negative time delay with respect to the reference transmission/reception time at the center of the array. The array elements will span the x and z axes of the local system, where the i-th element will have the following local coordinates $\boldsymbol{p}_i = [p_{i_x}, 0, p_{i_z}]^{\top}$. Accordingly, the generalized form of the i-th element of the receiver/transmitter steering vector given the relative local coordinate system and the g-th path is formulated as follows [95]:

$$\mathbf{a}_{j}(\boldsymbol{\psi}_{j_{e}}^{\prime},i) = e^{\frac{j2\pi}{\lambda}\boldsymbol{p}_{i}^{\top}\boldsymbol{u}_{j_{g}}^{\prime}}, \tag{26}$$

where $j \in \{\text{Tx}, \, \text{Rx}\}\$, $\lambda = c/f_c$ is the wavelength of the carrier frequency, and $u'_{jg} = u(\theta'_{jg}, \varphi'_{jg})$. Note that (26) can be reduced to a 2D coordinate system by only accounting for θ'_g and by reducing $p_i = [p_{ix}, 0]^{\top}$ and $u'_{jg} = [\cos(\theta'_{jg}), \sin(\theta'_{jg})]^{\top}$. The multiplication of the steering vectors $\mathbf{a}_{\text{Rx}}(\psi'_{\text{Rx}g})$ and $\mathbf{a}_{\text{Tx}}^{\top}(\psi'_{\text{Tx}g})$ results in a matrix of phase shifts of the size $N_{\text{Rx}} \times N_{\text{Tx}}$. If either the transmitter or the receiver does not have access to multiple antennas (i.e., a multiple-input single-output (MISO), single-input multiple-output (SIMO), or single-input single-output (SISO) system), the respective $\mathbf{a}_{\text{Rx}}(\psi'_{\text{Rx}g})$ and/or $\mathbf{a}_{\text{Tx}}(\psi'_{\text{Tx}g})$ should be eliminated from the equation. Hence, the matrix will reduce to a vector in the case of a MISO or SIMO system or be totally removed in the case of a SISO system.

As can be seen in (20)-(26), the FR2 channel model comprises elements that are directly related to the UE's position, velocity, and orientation, such as τ_0 , d_0 , v_0 , ψ'_{Tx_0} , and ψ'_{Rx_0} . Additionally, it contains elements that are related to the environment between the UE and the BS, which are captured in the multipath components of τ_g , d_g , v_g , ψ'_{Tx_g} , and ψ'_{Rx_g} , where $g \neq 0$. Hence, the accuracy of localizing the UE and mapping its surroundings is heavily dependent on the accuracy of the channel estimation algorithms used as well as the accuracy of the algorithms utilized to extract the positioning and mapping elements out of the channel response. Such algorithms will be discussed in the next subsection. In cases where FR1 is utilized, the bandwidth and number of antennas are limited compared to FR2, making the various paths' resolvability considerably challenging in complex environments [94]. In that case, stochastic channel models like Rician, Rayleigh,

and Nakagami distributions can be used to model small-scale fading caused by multipath. To overcome the limited localization accuracy obtained in such scenarios, these channel models can be coupled with RFs fingerprinting and AI-based positioning techniques to coarsely estimate the position of the user [96].

B. Ground Targets: Design Aspects

Over the past years, research endeavors in 5G TN positioning have exhibited substantial growth, encompassing a diverse spectrum of topics and methodological approaches. At a more abstract level, the domain of 5G positioning investigation can be partitioned into three principal domains: the estimation of 5G positioning measurables, the optimization of communication infrastructures to augment the quality of 5G positioning measurables, and the formulation of 5G positioning algorithms. Within each of these domains, researchers are engaged in furthering an array of sub-disciplines with the overarching aim of augmenting the precision and reliability of 5G positioning systems [79]. Before delving into the individual fields, it is worth noting that researchers in all of the aforementioned fields tend to derive error bounds for the various stages of the 5G positioning system. For instance, researchers involved with estimating the 5G positioning measurables will derive CRLB for these measurables to measure the optimality of their methods. On the other hand, researchers concerned with optimizing 5G systems for positioning purposes utilize these bounds as a theoretical framework to formulate their optimization problems [97]. Finally, researchers who developed 5G positioning algorithms utilize intermediate measurable bounds as well as the position error bound and the orientation error bound to provide an uncertainty quantification for their algorithms. Additionally, they utilize position and orientation error bounds as means to benchmark the optimality of their methods [40].

- 1) Estimation of 5G Positioning Measurables: The first research area focuses on estimating various measurables used in 5G positioning, including time-based and angle-based measurables and channel state information (CSI) [72,98,99]. The accuracy of these estimation algorithms has a direct impact on the accuracy of the positioning estimate. Additionally, within this domain, some researchers are interested in estimating the time synchronization bias between the UE and the BS [100]–[102] and the detection of NLOS signals [103]–[106]. Estimating synchronization bias would enable the use of TOA, reducing the reliance on TDOA and RTT measurements. NLOS signal detection is vital for improving the accuracy of positioning algorithms that solely depend on LOS signals. A specific aspect of NLOS detection algorithms is identifying the number of signal bounces encountered by the NLOS signal [107]. This classification is valuable for algorithms relying on NLOS signals for positioning, as they primarily utilize singlebounce reflections [108].
- 2) Optimization of Communication Systems to Enhance 5G Positioning Measurables: The second domain of research is centered on enhancing the design and performance of the communication system to improve the quality of measurements

used in 5G positioning. Subcategories within this domain encompass strategies like optimizing BS placement [41,109], enhancing power allocation and beamforming [97,110,111], and refining signal design [112]-[114]. Optimizing BS placement entails identifying the most advantageous BS locations within the network to enhance measurement quality, often by minimizing the GDOP. Power allocation and beamforming optimization involve adjusting the transmission power of BSs and beam direction to enhance measurement quality [97]. These approaches consider the utilization of both LOS and NLOS signals, optimizing power allocation per beam or direction [97]. Signal design optimization aims to create new signals tailored for positioning, like PRS and SRS, which can yield more precise measurements through improved autocorrelation properties [112]-[114]. Additionally, some designs seek to achieve higher signal-to-noise and interference ratios, further enhancing measurement performance.

3) 5G Positioning Algorithms: The third research area comprises the development and enhancement of positioning and localization algorithms that rely on 5G measurements. This field is primarily divided into LOS positioning and other minor subfields. In each sub-field, both snapshot and tracking algorithms are present, albeit more attention has been given to snapshot techniques. As the name suggests, LOS 5G positioning refers to the use of LOS 5G measurables and their uncertainties⁶ to determine the position of the UE that is either standing still or in motion. These algorithms can be categorized based on the measurements they use. Trilateration algorithms utilize TOA, TDOA, and RTT measurements from multiple BSs [115]-[117], triangulation algorithms utilize AOA and AOD measurements from multiple BSs [118], and hybrid algorithms utilize a mix of both range and angle measurements [101,119]-[125]. All LOS algorithms assume that LOS signals are available, detectable, and resolvable (i.e., from NLOS/multipath components). Such an assumption is valid, particularly when considering the wide bandwidth available in 5G NR, as there are other works that tackle such a problem, as mentioned in IV-B1. Nevertheless, there are certain assumptions in the literature that require careful attention from a practical point of view, as detailed in the following.

- *Tight BS-UE synchronization* is practically challenging, costwise, as time synchronization bias is expected to exist between the BSs and the UE due to the use of low-cost local oscillators [95]. Hence, direct utilization of TOA measurements without estimating such bias will be erroneous. One solution to address this issue is to jointly estimate the time synchronization bias and the position of the UE, as seen in [126]–[128].
- Constant connection to multiple BSs is challenging to realize in the communication-focused network deployment, where a single BS suffices. A more practical approach would be to follow the standards stipulated by 3GPP with regard to micro BS deployment in urban scenarios [92]. Additionally, utilizing quasi-real simulations where 3D maps of real-world

cities are utilized to dictate LOS and NLOS operations helps with realizing a more accurate testing scenario [79,129]. The distributed ultra-dense deployment of BSs, as in the case of *cell-free* networks, offers a promising solution to tackle this challenge (cf. Section VII).

Minor 5G positioning sub-fields include NLOS positioning, RFs fingerprinting techniques, cooperative (co-op) positioning, and integrating 5G with other sensors.

- NLOS positioning leverages NLOS and, if available, LOS measurables that may include reflections, diffractions, and scattering to estimate the position of the UE [130]–[134]. To do that, these methods tend to also co-estimate the position of the scatterers surrounding the UE/BS. Such methods are known as SLAM techniques and are rigorously investigated in the radar-, LiDAR-, and vision-based positioning literature [14].
- RF Fingerprinting techniques are mainly utilized in indoor scenarios, where the environment is less dynamic [135]– [137]. 5G indoor RFs fingerprinting methods utilize both RSS and CSI measurements/prints to estimate the UE location through various end-to-end machine learning (ML) techniques [5]. Such techniques perform poorly in outdoor settings, as the environment is more dynamic compared to indoor settings.
- Cooperative (co-op) positioning techniques are concerned with exploiting vehicle-to-vehicle communications to share data and positioning measurables to enhance the accuracy and the robustness of the positioning solutions [138]–[141].
- Sensor fusion of 5G and external sensors like GNSS, inertial
 measurement units (IMUs), and perception-based systems
 have been established in [79,142]–[144]. Such integration is
 essential to realize a seamless positioning solution during
 inevitable 5G outages. It is worth noting that this promising
 research area is expected to garner more attention as realworld 5G-based experimental setups/datasets start to become
 readily available for researchers around the world.

C. Aerial Targets: Special Considerations

This subsection focuses on the special considerations associated with positioning aerial targets within 5G terrestrial networks in an G2A localization system. Aerial UE poses unique challenges due to geometrical factors such as limited SNR caused by downward-tilted BS panels, poor GDOP depending on BS deployment, and the requirement for 2D arrays for altitude and attitude estimation [145,146]. Moreover, special considerations should be given to radio aspects like the reduced multipath effects and the potential applicability of FR1 for positioning purposes [147,148]. Finally, aerial targets will face challenges arising from high mobility, such as the need for frequent positioning updates and low-latency communications. Accordingly, this subsection is further divided into three aspects, namely, geometrical, radio, and mobility aspects.

1) Geometrical Aspects: In 5G terrestrial networks, BS panels are often tilted downward to optimize ground coverage, leading to limited SNR for aerial targets [145]. This reduced SNR can adversely affect aerial user equipment's positioning

⁶Uncertainties help with optimal weighting of the measurables. Lower uncertainty measurables should have higher weight compared to others, and vice versa.

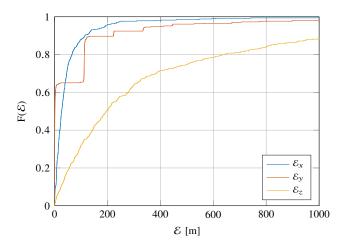


Fig. 6: An example CDF of TDOA-based aircraft localization error (in meters) in 3D along the horizontal (x and y) and vertical (z) axes using terrestrial anchors. The figure demonstrates the poor localization performance on the vertical axis compared to horizontal axes due to the high vertical GDOP. Target aircraft are at altitudes ranging from 8 km to 12 km.

accuracy and reliability. Additionally, the positioning accuracy of aerial targets can be influenced by the deployment of BSs, resulting in poor GDOP [146]. This is especially the case for altitude GDOP. Since aerial UEs are expected to navigate in 3D, information about horizontal and vertical relative angles is paramount. Hence, 2D antenna arrays at the BSs will be necessary to compute the 3D position and orientation of aerial UEs. In order to accommodate the relatively high speed and long distances aerial target travel, compared to ground targets, large-scale deployments, e.g., tens or thousands of kilometers, of ground anchors are needed to attain sufficient localization coverage for the aerial segment [146,149]. One type of network that can achieve such large-scale coverage is crowdsourced wireless networks, which rely on off-theshelf massively distributed software-defined radios to provide large-scale localization coverage. OpenSky-network [149] is one example of an established crowdsourced wireless network initiative, covering the majority of western Europe, focusing on aircraft monitoring by capturing ADS-B broadcasts. By capturing a given ADS-B message at four anchors, one can use TDOA-based MLAT to localize the corresponding aerial target [30]. While a crowdsourced wireless network offers large-scale localization coverage, such large-scale deployment makes the anchors synchronization needed for TDOA a challenging task [59]. Machine-learning-based clock models, such as LSTM (Long Short-Term Memory), can be used to predict and compensate for clock offset in crowdsourced wireless networks, boosting the accuracy of TDOA-based MLAT [30]. Nonetheless, due to the fact that ground anchors are majorly deployed into a 2D plane, the poor vertical GDOP remains an open challenge for aerial target localization with ground anchors [30,146]. Fig. 6 presents an example CDF of TDOAbased aircraft localization error along the horizontal x and y axes and the vertical z axis, respectively corresponding to \mathcal{E}_x , \mathcal{E}_y , and \mathcal{E}_z . In this figure, we consider four synchronized ground anchors aiming to localize a set of aircraft via TDOA measurements. Each anchor measures the TOA of the aircraft's ADS-B messages in nanoseconds. Subsequently, these measurements are fused using MLAT based on TDOA to provide a snapshot position estimate followed by a KF to enable aircraft trajectory tracking as detailed in [30]. As the figure demonstrates, the error along the vertical direction is noticeably larger, which is mainly attributed to the poor GDOP in the vertical dimension.

- 2) Radio Aspects: Unlike ground-based UEs, aerial targets experience fewer multipath effects, which imposes reduced requirements in terms of measurement resolvability [147]. This advantage is attributed to the limited bandwidth allocated for FR1 as well as the geometry of the environment. Therefore, the channel for aerial UEs using FR1 is expected to be narrowband with multipath following a Rician distribution [148]. Furthermore, the usage of FR1 will provide wider coverage compared to FR2, increasing the availability of the positioning solution. Hence, positioning using FR1 may hold greater promise compared to FR2, which is more suitable for ground UEs.
- 3) Mobility Aspects: Due to the higher dynamics and more stringent requirements of aerial targets, frequent positioning updates and low-latency communications are essential. The rapid movement of aerial UEs necessitates real-time positioning information. High-speed movement of aerial targets induces a high Doppler spread [150], resulting in challenges for positioning accuracy when multiple BSs are involved. Additional processing and advanced algorithms are required to compensate for the Doppler effect and maintain reliable positioning in multi-BS scenarios [151].

D. Key Takeaways

This segment delved into localization with ground anchors. While ground anchors have traditionally been utilized for indoor localization, recent advancements in cellular technology, particularly 5G, are extending this capability to outdoor spaces. Within terrestrial network positioning, research spans three primary domains: estimation of positioning measurables, optimization of communication systems, and development of positioning algorithms. Efforts in each domain focus on minimizing error bounds, optimizing system components, and devising algorithms for trilateration, triangulation, and hybrid positioning solutions. Open terrestrial localization challenges include mitigating synchronization bias (time and frequency), detecting LOS and NLOS operation, using the appropriate localization technique for each situation, and signal design to name a few. Furthermore, advances in 5G communications promoted niche research areas within the 5G terrestrial localization field, including NLOS positioning, RF fingerprinting, cooperative positioning, and sensor fusion with onboard motion sensors. In order to realize practical localization solutions, some pitfalls to be avoided. These pitfalls include assumptions about tight BS-UE synchronization and constant simultaneous connection to multiple BSs. Localizing aerial targets using terrestrial networks brings unique challenges caused by the limited SNR, poor GDOP, and high mobility, highlighting the need for frequent positioning updates and low-latency communications. Geometrical and radio aspects also play crucial roles, requiring 2D antenna arrays, large-scale ground anchor deployments, and optimization for frequency ranges to achieve wider coverage and reduced multipath effects.

V. LOCALIZATION WITH AERIAL ANCHORS

The integration of LAP and HAP in future wireless networks is becoming a reality, offering swift and on-demand connectivity for mobile users [145], as well as data collection [152] and wireless power transfer [153] for IoT devices. This integration at the RF communication level paves the way to exploit UAVs as aerial anchors for localization purposes. While Section IV elaborated on the techniques used by ground anchors to handle the multipath wireless channel, exploiting LOS and NLOS links individually, in this section, we focus on aerial anchors, which enjoy the luxury of influencing their wireless channel by leveraging their flexible placement and maneuvering both on the horizontal and the vertical dimensions. This flexible placement of aerial anchors unlocks the possibility of designing a dynamic radio localization system with adaptive anchor positioning. This section covers the main aspects of aerial-anchor-based localization systems designed to localize ground and aerial targets.

A. System Model

Designing a localization system using aerial anchors shares several similarities with its ground counterpart in terms of the signal strength, time, or angle measurements needed, the primary methods used to estimate the target position, and the filtering employed for target tracking. However, aerial-based localization systems introduce crucial new design aspects, such as adaptive 3D placement and trajectory design. In the following, we detail key system models and design aspects of aerial-based localization systems.

1) Radio Technology: The RF radios used in modern UAVs can be broadly categorized into ATM and C&C radios and data communication radios. A summary of the main radio technologies used in modern UAVs for both categories is presented in Table VI. In the category of ATM and C&C, the radios used mainly depend on the UAV type and, by extension, on the availability of visual LOS, cost, range, and reliability requirements. For instance, in rotary-wing UAVs, which typically fly at low altitudes (< 500 m), a combination of remote control and Wi-Fi are typically employed, whereas in HAPs and fixed-wing UAVs, which tend to fly at high altitudes (> 500 m), long-range protocols, such as ADS-B or Flarm, are used [3]. In the category of data connectivity, the types of radios carried onboard depend on the UAV mission's application, which could be a data collection mission form IoT nodes, e.g., using LoRa, a video streaming mission using LTE or 5G NR, or a connectivity mission in which the UAV acts as an aerial base station [148,154,155]. The radio technology mounted onboard a UAV is selected considering the tradeoff between data rate and range as well as constraints set by the application on reliability, cost, and security. Similar to the case of terrestrial anchors, the radio technologies carried onboard the UAV have a direct influence on the localization system design in terms of available bandwidth, power, latency,

TABLE VI: Range and transmission rate of the main RF technologies used in UAVs for both ATM and C&C as well as data.

	Technology	Max. Range	Message/Data Rate
	ADS-B	100 km	2 msg/s
	Flarm	10 km	0.1 msg/s
l margarithment [S	APRS	10 km	0.2 msg/s
M 85 .	Wi-Fi SSID	800 m	20 msg/s
ATM and C&C [3,158]	Cellular	2 km	20 msg/s
,	RC	1 km	25 msg/s
	Wi-Fi 2.4 GHz	500 m	450 Mbps
	Wi-Fi 5 GHz	50 m	1.2 Gbps
Data (7,154]	LTE	1 km	100 Mbps
Da 3,7,	5G FR1	1 km	100 500 Mbps
	5G FR2	500 m	2 Gbps
	LoRa	10 km	50 kbps

and the frequency used. The same holds when the UAV is the target to be localized. In that case, whether the UAV target is cooperative or non-cooperative, its message/data rate directly relates to the RF-based position update rate. It is worth noting that existing ATM and C&C radios in UAVs are mainly focused on communication, location updates, and synchronization with ground stations. However, in order to realize the foreseen dense deployments of UAVs, control messages for time synchronization, location updates, and interference coordination between UAVs need to be standardized. A UAV ATM ecosystem is under development aiming for standardized autonomously controlled operations of UAVs beyond visual LOS, which is led by FAA and NASA in the USA [156], and in parallel by SESAR (Single European Sky ATM Research) in Europe [157].

2) Channel Model: As discussed in Section III-B, a major source of error that hampers the quality of localization measurables, in general, is the randomness resulting from the shadowing effect [54]. Aerial anchors can swiftly adapt their 3D position, which has a direct impact on the LOS probability and, hence, on the shadowing effect. In order to model the largescale fading channel between an aerial anchor, e.g., a UAV, at an altitude h above ground level, and a ground target, an altitude-dependent pathloss model is adopted in [36,159]. This pathloss model represents the received power as a function of the direct distance d and the elevation angle $\varphi = \tan^{-1}(h/r)$. This model follows the formulation introduced in (11) to superimpose pathloss and shadowing; however, it is revised to incorporate the dependencies of shadowing and the pathloss exponent on the elevation angle, as presented in [160]-[163]. Accordingly, the pathloss exponent n_p and shadowing effect X in (11) can be respectively replaced by $n_p(\varphi)$ and $X(\varphi)$. Here, $n_p(\varphi)$ is the elevation-angle-dependent pathloss exponent and $X(\varphi) \sim \mathcal{N}(0, \sigma_X^2(\varphi))$ represents the elevation-angledependent shadowing effect which is a normally distributed random variable with zero mean and variance $\sigma_X^2(\varphi)$, in dB. The elevation-angle-dependent standard deviation $\sigma_X(\varphi)$ with an independently distrusted LOS and NLOS, is given as [160]

$$\sigma_{X,j}(\varphi) = a_j \exp(-b_j \varphi), \quad j \in \{\text{LOS}, \text{NLOS}\},$$
 (27)

with a_j and b_j being frequency and environment-dependent parameters. Now, the overall average shadowing effect in the

links can be represented by the variance written as

$$\sigma_{X}^{2}(\varphi) = \mathcal{P}_{\text{LOS}}^{2}(\varphi) \cdot \sigma_{X, \text{LOS}}^{2}(\varphi) + [1 - \mathcal{P}_{\text{LOS}}(\varphi)]^{2} \cdot \sigma_{X, \text{NLOS}}^{2}(\varphi),$$
(28)

where \mathcal{P}_{LOS} is the probability of LOS. Equation (28) implies that the shadowing effect gradually diminishes with increasing aerial anchor's altitude due to the decreased probability of encountering obstacles between a transmitter and a receiver [148,162]. The LOS probability can be calculated empirically based on measurements [164], deterministically based on ray tracing [129], or stochastically based on geometry [165].

In terms of small-scale fading in UAV communication channels, the Rician model is widely adopted since it accommodates LOS, which has a relatively high probability, and multipath fading [151]. In Rician fading, the Rician K factor is a quantitative parameter to represent the severity of the NLOS multipath fading. UAV altitude-dependent and elevation-angledependent Rician K factors are considered in the literature for A2A channels [166] and A2G [148], respectively. Moreover, UAV channels tend to possess a higher Doppler effect compared to their terrestrial counterparts because the relative velocity and mobility of UAVs are higher [151]. Large Doppler spread occurs when the transmitting UAV is relatively close to the ground or aerial user. However, suppose the transmitting UAV is further away, e.g., at a sufficiently high altitude. In that case, multipath components will experience a similar Doppler frequency since reflective objects in close proximity to the receiving user will all be seen under similar angles from the transmitting UAV. Such Doppler effect can be mitigated via frequency calibration and synchronization [147].

3) Anchor Placement: UAVs' versatility and maneuverability enable flexible 3D placement, which is arguably the main distinguishing feature UAV aerial anchors have over their terrestrial and celestial counterparts. Since aerial anchors are mobile in nature, their placement refers to trajectory design. To fully exploit the benefits offered by this distinctive characteristic, extensive research efforts focused on trajectory design and optimization of UAVs integrated into wireless networks [167]. In particular, several studies addressed the trajectory optimization of UAV anchors, aiming at leveraging their flexible positioning in designing aerial-based localization systems [36,37,85,168]. To obtain a tractable form of the UAV trajectory with a finite number of optimization variables, path discretization is widely adopted in the literature [169], in which the UAV path is discretized into L line segments represented by L+1 waypoints. The length of a line segment is chosen, taking into account mission characteristics such as UAV velocity, total mission duration, and area of interest. Subsequently, a UAV trajectory can be characterized by a set of waypoints with a known size and the duration the UAV spends flying sequentially between waypoints. The set of waypoints is denoted by $\{\mathbf{w}_l\}_{l=0}^L$, where $\mathbf{w}_l \in \mathbb{R}^3$ is the 3D coordinates of the *l*-th waypoint. The duration the UAV spends within each line segment is denoted by $\{T_l\}_{l=0}^{L-1}$, together indicating a trajectory $\mathcal{T}(\{\mathbf{w}_l\}, \{T_l\})$. It is worth noting that the path discretization presented here can also describe cases where a rotary-wing UAV hovers at a given location by having a line segment with a length equal to zero, i.e., $\mathbf{w}_l = \mathbf{w}_{l+1}$. A generic

formulation for UAV trajectory design in an aerial-based radio localization system can be expressed as

minimize
$$\mathcal{T}(\{\mathbf{w}_l\}, \{T_l\})$$
 (29a)
subject to $f_i(\{\mathbf{w}_l\}, \{T_l\}) \leq \gamma_{f_i}, i = 1, 2, \dots,$ (29b)

$$p_i(\{\mathbf{w}_l\}, \{T_l\}) \le \gamma_{p_i}, \ i = 1, 2, \dots,$$
 (29c)

$$g_i(\{v_l\}) \le \gamma_{g_i}, i = 1, 2, \dots,$$
 (29d)

$$E \le \gamma_E,$$
 (29e)

where $U(\mathcal{E}_{\mathbf{p}})$ is the objective function, which could describe the error of a single target or multiple targets averaged in the time or the spatial domain. The constraint functions (29b)-(29e) are, respectively,

- f_i models communication-related constraints such as spectral efficiency, resource allocation, latency, as well as coverage, each represented by a threshold γ_f .
- p_i represents localization-related constraints, which include the targeted GDOP, type of measurements (e.g., RTT needs more time to conduct than TOA), LOS/NLOS situations, and the number of waypoints per UAV (e.g., four waypoints if a single UAV is used).
- g_i models mobility-related conditions as a function of the segment speed $v_l = (\mathbf{w}_{l+1} \mathbf{w}_l)/T_l$, $\forall l$, such as the maximum speed constraint of UAVs and the turning radius constraint upper bounded by the corresponding threshold γ_{g_i} .
- E defines the total energy consumption of the UAV anchors bounded by a limit γ_E .

The total energy consumption of a UAV is generally a sum of two components: the propulsion energy, which is required to ensure that the UAV remains aloft as well as to support its mobility, and the communication-related energy used for radiation/reception and signal processing. The propulsion energy depends on whether the UAV is fixed-wing [170] or rotary-wing [85,169] and is typically expressed as a function of the UAV's instantaneous velocity. The communication-related energy mainly depends on the wireless technology onboard. For instance, in large-size UAVs that are deployed in remote areas to collect information from IoT nodes, the communication-related can be neglected, whereas in small-size UAVs that carries a broadband communication technology onboard, the communication-related energy must be considered in the optimization problem.

Optimizing UAVs' trajectory using (29) minimizes the localization error and improves the localization accuracy. However, the tradeoff between this better accuracy, on the one hand, and the localization coverage and latency, on the other hand, must be carefully analyzed [85]. The localization coverage aspect is relevant whether (29) optimizes the trajectory of single or multiple aerial anchors. Since the communication coverage changes as the UAV travels from one waypoint to another [148], it is important to bound the coverage to the area where reliable localization services are guaranteed. The latency aspect is more relevant to the case when a single UAV is used. In this case, a position fix is only possible after collected measurements at several waypoints, e.g., \geq 4 for 3D positioning, enabling a single UAV to act as multiple virtual

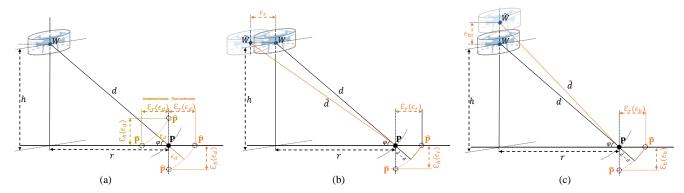


Fig. 7: Ranging error bounds with aerial anchors caused by (a) range underestimation or overestimation errors, (b) anchor displacement error on the horizontal plane, resulting from GPS error, (c) anchor altitude error, resulting from barometer error. The point W denotes the scheduled waypoint (the position where the UAV is supposed to be), and the point \tilde{W} denotes the actual position where the UAV is present.

anchors. The time needed to collect these measurements is bounded by the UAV mobility constraints such as maximum forward and turning speed. Accordingly, when optimizing an aerial anchor's trajectory, coverage and latency requirements can be incorporated into (29b) and (29d), respectively.

4) Anchor Placement Inaccuracy: Considering inaccuracy in anchor locations is particularly crucial in aerial-based localization systems. Unlike terrestrial anchors, which are typically stable on the ground, aerial anchors are exposed to weather conditions such as wind and rain, directly influencing their hovering and cruising stability. This weather influence causes a rapidly varying true anchor position, which cannot be accurately captured by the onboard barometer and the GNSS receiver, introducing an error in the local position estimate of the aerial anchor. Consequently, the ranging error between the aerial anchor and the target is a function of three independent components, namely, the direct distance estimation error ϵ_d , the anchor's horizontal displacement error ϵ_r , and the anchor's altitude error ϵ_h [171]. In Fig. 7a, 7b, and 7c, we visualize the effect on horizontal and vertical ranging error bounds caused by the total ranging error, horizontal UAV displacement error, and error in UAV's altitude, respectively. By applying geometric rules, one can derive the ranging error bounds for a target in 3D both on the horizontal xy-plane as well as the vertical plane. The maximum ranging error projected on the xy-plane for a given ϵ_d , ϵ_r , and ϵ_h can be approximated as [171]

$$\mathcal{E}_r(\epsilon_d, \epsilon_r, \epsilon_h) \approx \epsilon_r + \frac{h}{r}\epsilon_h + \epsilon_d \sqrt{1 + \frac{h^2}{r^2}},$$
 (30)

where h is the anchor's true altitude, and r is the true distance projected on the horizontal plane between the anchors and the target. Following the derivation of ranging error bounds projected on the horizontal plane [171], the maximum ranging error projected on the vertical plane can be expressed as

$$\mathcal{E}_h(\epsilon_d, \epsilon_r, \epsilon_h) \approx \epsilon_h + \frac{r}{h}\epsilon_r + \epsilon_d \sqrt{1 + \frac{r^2}{h^2}}$$
 (31)

Note that (30) and (31) are sufficient to account for anchor placement error in case ranging-based localization is adopted, e.g., RSS-based or time-based. However, in the case of angle-based localization such as AOA or AOD, the orientation

error of the UAV should also be considered. In the following subsection, a simulated ranging example is presented to study the effect of anchor position inaccuracy, considering different values of ϵ_h and ϵ_r .

B. Ground Targets: Design Aspects

In this section, we focus on aerial anchors for ground target localization, namely A2G localization. The research fields on ground target localization using aerial anchors mainly focus on exploiting adaptive aerial anchor placement. These fields can be broadly categorized into two fields, namely, enhanced localization measurables and localization algorithm design.

1) Enhanced Localization Measurements: The randomness of the wireless channel resulting from shadowing, a.k.a. large-scale fading, and multipath, a.k.a. small-scale fading, is generally considered as a critical source of error in radio localization systems [54]. In A2G localization systems, where aerial anchors are employed, the vertical dimension can be exploited to minimize the shadowing effect and improve the LOS experience (cf. Section V-A2), affecting the localization performance positively [36,172].

UAVs favorable channel conditions due to their higher altitude triggered several recent works to investigate RSSbased localization as a low-cost and simple solution for widerange outdoor wireless networks [36,37,85,168]. In [36], the positioning of multiple UAVs flying at the same altitude is optimized with the objective of minimizing the average localization error with IoT nodes being the targets to localize. The results showed that UAVs' altitude plays a significant role in the localization performance, and this positive effect is more pronounced in urban areas compared to suburban areas. Fig. 8 presents the impact of aerial anchors' altitude on RSS-based ranging accuracy. In this figure, we used a Monte Carlo simulation to study the effect of a UAV's altitude, h, on the average ranging error of ground IoT nodes uniformly distributed in an urban area with a radius of 1000 m. RSSbased ranging is done by incorporating $n_p(\varphi)$ and $\mathcal{X}(\varphi)$ into (11) and solving the resulting equation for d, as detailed in [36]. The figure shows that an optimal altitude exists at which aerial anchors considerably outperform their terrestrial counterparts. The decreasing trend of the average ranging

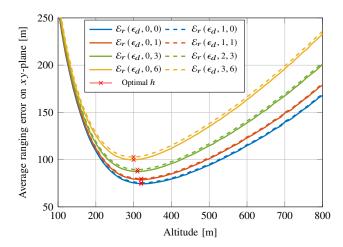


Fig. 8: Average ranging error projected on the xy-plane versus aerial anchor's altitude h, considering distance estimation error as well as aerial anchor's position inaccuracy.

error at low altitudes is attributed to the decreasing shadowing effect's variance. The increasing trend of the average ranging error is mainly due to the diminishing slope of the RSSdistance logarithmic curve at high altitudes, making distance estimation significantly more sensitive to any relatively small variations in the RSS measurements. The error resulting from the inaccuracy in the aerial anchor position when localizing ground nodes is studied in Fig. 8 by using the projected ranging error bound defined in (30). As shown in the figure, the ranging error projected on the ground is more sensitive to the vertical error in the anchor's position ϵ_h compared to the anchor's horizontal displacement error ϵ_r . This sensitivity to the anchor's vertical error ϵ_h does not only introduce a ranging error, but it also influences the optimal altitude. In addition to raining accuracy gains, results in [36] presented the benefits in terms of the required number of anchors, demonstrating that a lower number of aerial anchors are required to meet a given localization accuracy compared to localization solutions with terrestrial anchors.

The use of time-based localization methods with aerial anchors is investigated in [171]-[173]. Unlike RSS-based localization methods, which are mainly influenced by the propagation environment, time-based methods are also affected by the anchor's hardware, such as the local clock stability and the supported bandwidth. In the presence of a stable local clock and sufficient bandwidth, time-based localization methods are able to outperform RSS-based methods by orders of magnitude [171,173]. In [171], the authors demonstrated that by using TOA, a localization accuracy in the scale of submeters can be achieved when choosing an elevation angle, φ , that maximizes the LOS probability. However, the direct distance, d, was limited to a maximum of 60 m, which is relatively low for aerial anchors. To address the A2G localization problem in NLOS environments, the authors in [173] introduced an NLOS bias into the TOA measurement model and proposed an algorithm that estimates the average bias together with target location to improve localization accuracy. Reported results showed a noticeable localization performance gain, with the tradeoff being the computational complexity. The use of hybrid TOA/AOA measurements in aerial anchors is investigated in [174], where LOS link of aerial anchors were exploited to localize targets in 3D space.

2) Trajectory Optimization: As discussed in Section V-A3, in addition to the waypoint coordinates $\{\mathbf{w}_l\}_{l=0}^L$, UAV's trajectory in 3D, $\mathcal{T}(\{\mathbf{w}_l\}, \{T_l\})$, is characterized also by the number of waypoints, L+1, and the time spent flying between and at them, $\{T_l\}_{l=0}^{L-1}$, [37,85]. In [85], an energy-constrained trajectory design of rotary wings UAVs is explored, aiming to minimize the localization error of ground IoT nodes. The reported results showed that a mobile UAV anchor could outperform ground anchors in terms of localization performance. While the results demonstrated the major effect of optimizing the UAV anchor's altitude, it was also shown that when the hovering time at waypoints is relatively low, trajectory optimization on the horizontal dimension is also crucial. In [37], the authors introduced a reinforcement-learning-based framework to minimize the RSS localization error of ground nodes by optimizing the UAV trajectory, considering operational time, number of waypoints, trajectory shape, velocity, altitude, and hovering time. The results showed that there is a tradeoff between the localization error and path length. In particular, increasing the UAV's hovering time, number of waypoints, and velocity remarkably improves the localization performance at the cost of a longer path or longer operation time and hence, higher energy consumption. In [168], the authors capitalized on the environment's 3D map to optimize the UAV trajectory to localize terrestrial users using RSS measurements under a given mission duration.

Apart from ranging-based A2G localization, the mobility of aerial anchors can be exploited in designing non-MLAT range-free localization algorithms [171]. The essence of such range-free localization algorithms is the heard and not-heard (HnH) technique, in which a stationary ground target exploits a periodically broadcast signal from an aerial anchor moving in a given trajectory to define an area where it may reside and places itself at the center. The broadcast signals from the aerial anchor include its current position (i.e., waypoint). Examples of such range-free algorithms are drone range-free (DRF) [175] and the intersection of circles (IOC) [55]. In general, range-free algorithms suffer from poor localization accuracy due to their dependency on the antenna radiation pattern quality [176]. In [171], the authors extended DRF and IOC algorithms to include TOA-based distance measurements, resulting in localization error is 3-4 times lower than the original range-free versions.

C. Aerial Targets: Special Considerations

The A2A localization scenario, where aerial targets are localized using aerial anchors, which may include LAPs anchors and HAPs anchors, introduces several advantages and challenges when compared to localizing aerial targets using terrestrial anchors [177]. On the one hand, the 3D placement of aerial anchors offers a promising solution to the vertical GDOP, and links between aerial anchors and aerial targets are dominantly LOS. However, on the other hand, the high

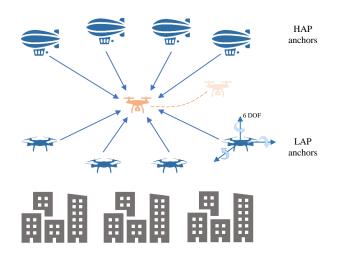


Fig. 9: A2A localization: Different altitudes of aerial anchors offer a promising GDOP experienced both from the horizontal and vertical perspective.

mobility of aerial anchors makes the Doppler effect more challenging, in addition to extra challenges related to the radio technology used, such as the 3D localization coverage and the need for relatively long-range communication. Aerial targets generally have high dynamics at relatively low altitudes (lower than 5 km). Such target include rotary-wing or fix-wing drones and other aircraft that can be regarded as LAPs [7]. While the need to localize HAPs can be neglected because of their quasi-stationary characteristics [178], they are highly expected to serve as anchors. In the following, we detail the key aspects that differentiate aerial targets from ground targets when using aerial anchors for target localization.

- 1) Geometrical Aspects: Unlike ground and space anchors, aerial anchors can be present at higher and lower altitudes with respect to the aerial target, offering a better vertical GDOP experience, as visualized in Fig. 9. While using joint LAPs and HAPs to localize aerial targets improves GDOP, it brings deployment and coordination challenges in terms of both horizontal distances and height differences among aerial anchors. The joint deployment of LAP and HAP networks is investigated in [179,180]. A general model of collaborative UAV networks and corresponding position error bounds were proposed in [181]. The authors investigated the effect of the number of UAVs and the use of direct or multi-hop links between them on localization accuracy. It has been shown that increasing the number of hops used to share coordinates and distance measurements degrades the localization accuracy. In order to minimize the deployment and operation costs of A2A localization systems, the localization coverage area can be confined in a given zone, e.g., secure no-fly zone. The use of surveillance UAVs for detecting and localizing aerial targets in a defined no-fly zone is explored in [43].
- 2) Radio Aspects: A2A links are relatively more clean than G2A links which experience serious multipath effects [182]. Thus, the almost pure LOS channel characteristics facilitate LOS-based localization using multiple aerial anchors, thanks to the decreasing trend in the cost of deploying massive UAV swarms. On the one hand, LOS links are beneficial for A2A

radio localization, aiding both accuracy and coverage. However, on the other hand, they make the interference problem significantly more challenging [183]. This tradeoff between localization coverage and interference will be particularly explicit in setups adopting relatively low-frequency long-range communications such as FR1. The use of the mmWave band can address the interference problem by relying on narrow directive beams, which also serve angle-based localization. However, the communication range will be determined by mmWave propagation loss, which may result in a reduced A2A localization coverage. Radio aspects in A2A localization scenarios should also consider whether LAPs, HAPs, or both are used as anchors [179,180]. For instance, HAPs can be used as anchors as well as to regularly update the distanceangle information among LAP anchors. Hence, long-range communication can be used for HAP-LAP links, whereas shorter-range communication can be used among LAPs in an energy-efficient manner.

3) Mobility Aspects: In addition to the geometry and radio aspects in A2A localization, we herein emphasize the mobility aspects represented by the tracking of aerial targets with aerial anchors. Unlike the G2A localization, the A2A localization is more likely to deal with the dual mobility of anchor and target. The 6D mobility, which geometrically includes 3D translational mobility and 3D rotational mobility, of rotarywing drones will induce many micro-Doppler components. Moreover, the placement and orientation of the antennas mounted on the aerial anchors play an important role in receiving signals from aerial targets due to the polarization and directivity. This antenna placement issue is more pronounced in A2A localization scenarios since aerial targets may be above or below the aerial anchor. The inherent 6D mobility of UAVs has an influence on angular-based localization. In [184], the authors utilized AOA for target localization, where the UAV six-DOF dynamic model is formulated, and a diffusion extended KF is used to implement distributed adaptive estimation. However, the trajectory optimization complexity of aerial anchors becomes very high when considering six-DOF in the angular domain. To address this complexity, one potential solution is to employ HAPs, exploiting their quasistatic behavior to minimize fluctuations in angle estimation.

D. Key Takeaways

Aerial anchors use radio technology both for data as well as for ATM and C&C. They enjoy adaptive on-demand placement, which can be optimized to improve the LOS experience and, hence, the localization performance. To optimize an aerial anchor's trajectory to minimize localization error, aspects to take into account include radio communication requirements, localization measurements type, anchor mobility, as well as anchor's energy budget. While several works in the literature addressed aerial anchors' trajectory optimization, solutions that jointly consider communication, localization, mobility, and energy aspects are still an open research direction. Despite its enormous advantages, this adaptive placement comes with a cost related to the fact that aerial anchor placement is prone to errors that translate into target localization errors.

One interesting open research direction is to study and model the key environmental factors that cause this error, such as wind speed and rain, combined with a KF to enable accurate placement calibration. Another potential solution may rely on harnessing a known reference point, such as a ground station, to act as a reference to calibrate aerial anchor placement. In addition to aerial anchors' advantages in ground target localization, their 3D placement enables addressing GDOP both horizontally and vertically in aerial target localization. However, the tradeoff is the excessive Doppler effect that may result from target and anchor mobility, particularly when considering fixed-wing aerial anchors.

VI. LOCALIZATION WITH SPACE ANCHORS

In recent years, the space industry has changed dramatically. The new space industry is dominated by commercial players, which is a break from the past when space was almost only accessible by governments. Launching satellites has become more affordable, especially launching satellites to LEO [15]. This has led to a large number of LEO telecommunication satellite constellations being proposed and launched. The major advantage of using LEO satellites in comparison with satellites in a higher orbit is the lower communication delay and lower pathloss [185]–[187]. On the downside, each satellite can only cover a smaller area of Earth's surface. As a result, a LEO constellation consists of a large number of satellites to reach global coverage. Therefore, LEO satellites are often small and built from commercial off-the-shelf components to reduce costs. This section explores localization using NTNs with space-based anchors. More specifically, we will focus on the use of LEO satellites for localization. LEO satellites have enticing attributes that are useful for localization purposes [15,188]. Current space-based localization is based on GNSS, which resides in MEO at around 20,000 km above the surface of the Earth. The shorter distance to Earth of LEO satellites leads to high signal power and a shorter delay. LEO satellites often use different frequency bands, leading to a higher frequency diversity, which hardens space-based localization against jamming and spoofing. In addition, the large constellation sizes and the wide variety of different orbits increase the geometrical diversity.

There are two types of LEO satellites that can be used for localization. The first is communication LEO satellites, which can be used as an opportunistic source of localization signals; this is called the signals-of-opportunity (SOP) method [189]. The main advantage is the reuse of already deployed infrastructure. However, using these proprietary systems for localization might not be trivial as the used technology might not be disclosed. The second type is the LEO-positioning, navigation, and timing (PNT) constellations, which are specifically designed to provide PNT services from low Earth orbit. Several commercial companies have launched initial LEO-PNT test satellites, for example, the PULSAR system from Xona Space Systems [190], the GeeSAT satellites from GeeSpace and the CentiSpace satellites from Future Navigation [191]. Governmental institutes are also interested in LEO-PNT. For instance, the ESA (European Space Agency) is planning to launch several test satellites in the near future as part of their FutureNAV program [192].

A. System Model

Localization using LEO satellites brings some challenges [193]. In order to localize targets, the location of the anchors has to be known. In the case of satellites, the anchors are in an orbit around Earth. As a result, the anchors are constantly moving, albeit in a very predictable way [15]. Still, the precise orbit needs to be known before the target can be localized, and errors in the orbit data result in localization errors. In addition, the stability of the onboard satellite clock is very important as this can lead to time and frequency errors [188]. For time-based ranging methods, a small time offset results in a sizeable ranging error, and Doppler ranging methods rely on precise frequency offset measurements. The next challenge is the signal structure of the communication signal. Localization methods often rely on searching for known signals. However, often, the used signals of commercial satellite communication systems are proprietary, resulting in difficulties in tracking the signals. In addition, the signals might not be optimized for being used on localization systems, resulting in sub-optimal performance. All of the aforementioned system aspects are discussed in the following.

1) Radio Technology: The first parameter that influences radio technology and its signal design is the frequency of operation. The chosen band has to be available following the frequency plan. For different applications, different parts of the spectrum are reserved. Furthermore, the chosen frequency dictates the available bandwidth; more bandwidth is available at higher frequencies. In addition, higher bandwidth is generally beneficial for localization performance as the time resolvability and measurement accuracy can be increased when a larger bandwidth is available [194]. In Table VII, we present an overview of the radio technology used in existing commercial LEO constellations. Note that the constellation sizes presented in this table are constantly changing as satellites are being launched at the highest rate ever in history. When designing the modulation of the signal, the ranging properties of the signal have to be taken into account. A ranging signal should have good auto-correlation properties. When correlating the signal with itself, the response shows a high peak when the signals completely overlap and no response in all other cases. The width of the peak depends on the symbol rate of the sequence, which in turn relates to the bandwidth of the signal. A narrow peak results in a higher ranging accuracy. GNSS signals use pseudo-random noise (PRN) codes, which have excellent auto- and cross-correlation properties. These PRN codes are modulated on the carrier using binary phase-shift keying or binary offset carrier modulations. However, when considering LEO communication satellites, the modulation also has to be optimized for communication purposes as well. Therefore, OFDM waveforms have to be considered. With the recent inclusion of the PRS in 5G networks, an excellent ranging signal is added to the OFDM-based standard [195]. The PRS is a downlink signal that can be scheduled on different resource blocks, resulting in a configurable frequency

Name	Use	Height [km]	Modulation	Frequency	Bandwidth	Proposed/Current size
Iridium NEXT	BB	780	PSK	1618.25 MHz	31.5 kHz	66 / 66
StarLink	BB	550	OFDM	10.7-12.7 GHz	250 MHz	11927 / 5636
Kuiper	BB	610	-	17.7-20.2 GHz	100 MHz	3236 / 0
OneWeb	BB	1200	OFDM	10.7-12.7 GHz	250 MHz	648 / 634
GlobalStar	Voice	1420	CDMA	1618 MHz	1.23 Mhz	42 / 25
Sateliot	IoT	505	NB-IoT	2000 MHz	180 kHz	250 / 0
OmniSpace	IoT	530	NB-IoT	1990 MHz	180 kHz	200 / 2
OQ Technology	IoT	480	NB-IoT	2005 MHz	180 kHz	10 / 10
Lacuna	IoT	500-550	LR-FHSS	915 MHz	336 kHz	240 / 6
Swarm	IoT	300-550	LoRa	149 MHZ	20,8-62,5 kHz	150 / 164
Myriota	IoT	530	GMSK	400 MHz	150 kHz	50 / 3
OrbComm	IoT	710	PSK	149 MHz	15 kHz	50 / 50

TABLE VII: Radio technology used in commercial LEO NTNs.

and time span. The used symbols are gold sequences with excellent auto-correlation properties. This signal is a good starting point for enabling pseudo-range measurement with space-based communication systems.

2) Channel Model: Several channel models have been developed for LEO communication satellites. However, there is not one complete channel model that covers all LEO constellations. These channel models depend on multiple factors linked to the LEO constellations considered, as well as the system's frequency and bandwidth. An overview of channel models is given in [196]. Therefore, in the following, we will discuss some of the key aspects that have to be taken into account in modeling satellite-to-Earth channels.

Due to the high mobility of LEO satellites, the Doppler frequency and rate are high. The Doppler frequency (i.e., Doppler shift) is highest for satellites with a low elevation angle (e.g., 10°), while it approaches zero for satellites a high elevation angle (e.g., right overhead at 90°). When acquiring and tracking LEO signals, taking the Doppler frequency into account is essential. Assuming a perfectly known CFO and received signal frequency f_r , from (17), the Doppler frequency is defined as $f_D := f_r - f_c$. It can be written as a function of the relative radial velocity Δv between the target and space anchor as follows

$$f_D = \frac{\Delta v}{c} f_c \,, \tag{32}$$

where f_c is the carrier frequency used by the satellite. For broadband networks using K-band frequencies ($\sim 12 \text{GHz}$), the Doppler frequency can be up to 250kHz [197].

When transmitting signals through the ionosphere, the signal experiences an ionospheric delay [198]. Due to solar radiation, the ionosphere is a partially ionized medium. The delay of the signal depends on the total amount of free electrons along the signal's path, noted as the slant total electron content (STEC) [198]. As this is influenced by solar radiation, the ionospheric delay is the largest during the day. The phase and the modulation of the signal experience a different delay. The modulated code gets delayed, called the *group delay* Δ_{group}^{iono} , while the phase actually experiences an advance Δ_{ph}^{iono} [199]. The size of the delay and the advance can be calculated as

$$\Delta_{ph}^{iono} = -\frac{40.3}{f_c^2} \cdot \text{STEC}, \quad \Delta_{group}^{iono} = \frac{40.3}{f_c^2} \cdot \text{STEC}.$$
 (33)

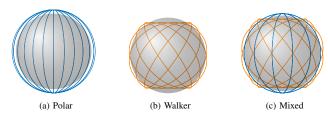


Fig. 10: Polar orbits have a high inclination angle ($\theta_i \approx 90^\circ$) and have global coverage. However, they provide the best coverage in polar regions. Walker constellation has a lower inclination angle ($\theta_i \approx 55^\circ$). They optimize the coverage over the population hotspots at mid-latitudes but lack coverage at the poles. A mixed constellation can provide the best of both.

The higher the frequency, the lower the ionospheric effect. Due to the frequency dependence, the ionospheric effect can be estimated when using two separate frequencies. For example, all GNSS constellations are transmitting ranging signals at multiple frequencies to be able to correct the ionospheric delay.

Higher frequencies require smaller antennas, as the size of an omnidirectional antenna is coupled to the wavelength of the carrier frequency. This can be an advantage as this makes the system more compact. However, the received power will also decrease due to the decreased antenna surface. As the distance between the transmitter and receiver is relatively large, increasing the received signal power is critically important. One popular solution exploited in satellites is the use of antenna arrays to create directional beams to increase signal power.

3) Anchor Placement: A satellite constellation is a group of satellites with the same function and with similar orbits. The way the orbits of the satellites are designed has a large influence on the coverage of the system. The most important parameters in constellation design are the orbit height, inclination angle, and constellation size [188]. A higher orbit has the advantage of being able to cover a larger area but increases the free space pathloss, propagation delay, and launch cost. The inclination angle is the angle between the orbital plane and the equatorial plane. The inclination greatly influences where on Earth the coverage is optimized [200]. Polar orbits have an inclination angle close to 90°. They require the least amount of satellites to reach global coverage. However, the coverage is highest in the polar regions, while the population density of Earth is not concentrated around the poles. Using a Walker constellation with inclined planes of 50° to 60° optimizes the coverage at the major population centers at mid-latitudes but lacks coverage at the poles, as depicted in Fig. 10. The selection of the constellation type depends on the system's use case. A combination of mixed polar and inclined orbits can bring the best global coverage while optimizing the coverage at population density hotspots. The third parameter to optimize is the constellation size. The amount of satellites needed to reach global coverage depends on how many satellites must be visible at the same time for a user. For localization purposes, a minimum of four visible satellites is needed. The number of satellites needed to reach the targeted coverage can be calculated depending on the chosen orbit height and inclination. The authors of [201] use a genetic algorithm to design a positioning constellation, optimizing for GDOP while minimizing constellation size.

Unlike aerial anchor placement, which can be controlled and adapted on-demand, space anchors follow predefined orbits that are prone to perturbations. For instance, effects such as atmospheric drag are particularly pronounced at LEO orbits. Therefore, even after designing the orbit height, inclination angle, and constellation size, orbit determination remains a key aspect of user localization. To localize users based on space anchors, the precise orbit of the satellites must be known. For LEO satellites, this can be achieved in two ways. The first method makes use of the already available GNSS constellations. A GNSS receiver is deployed in orbit on the LEO satellite. In this way, the precise orbit of the satellite can be determined. However, this makes the system reliant on GNSS, which means that if these constellations become unavailable, the orbit of the LEO anchors can not be determined. In addition, the accuracy of this method is limited unless the onboard GNSS receiver can be provided with correction signals needed to enable precise point positioning [15]. To acquire more precise orbit information, a ground monitor network is needed. Such a network consists of static receivers that are continuously monitoring the signals coming from the satellites. Based on the measurements of the network, a more precise location of the satellites can be estimated.

- 4) Clock Synchronization: To exploit LEO satellite signals as a ranging reference, the clocks of the different satellites have to be mutually synchronized. GNSS satellites use large, heavy, and expensive atomic clocks to keep a very accurate timing reference on board the satellite. As a typical LEO constellation usually exists out of a large fleet of small and inexpensive satellites, such atomic clocks are not an option. Alternative options include:
- Use smaller, less accurate atomic clocks on board the LEO satellite. As the satellites orbit the planet several times a day, the clocks can be adjusted using correction signals transmitted from a fixed position on Earth. In this way, the period in which the clock has to be stable is limited to a couple of hours.
- Exploit higher orbits satellites disciplined clock, e.g., GNSS disciplined clock. For this strategy, the satellite contains a GNSS module and an oven controlled crystal oscillator (OCXO) to keep the time. The GNSS module is used to receive the signals of the GNSS constellations and compute the time. This information is used to discipline the OCXO

- and correct the drift of the clock. In [202], the authors present how they used an Ultra Stable Oscillator in combination with a triple frequency GNSS receiver to achieve sub-nanosecond time synchronization.
- Employ optical inter-satellite links (OISLs) to enable communication and synchronization between the satellites in a constellation. OISLs make use of lasers deployed on the satellites to set up high-speed communication links with excellent ranging performance [203]. The authors of [204] show that employing OISLs in a GNSS constellation can increase the accuracy of the precise orbit determination and decrease the signal-in-space ranging error.

B. Ground Targets: Design Aspects

In this section, we focus on two aspects regarding the usage of LEO satellites for localization purposes. In particular, we discuss the various design aspects of LEO-based localization systems and the opportunistic usage of SOP from communication-based LEO satellites. The design of space-based localization requires a careful trade-off between accuracy, cost, power, weight, and size, as discussed in the following.

- 1) Constellation Selection: When using NTN LEO constellations for localization, constellation selection can be done based on use-case. Table VII lists the major operational and proposed communication satellite constellations along with their corresponding localization use cases in the literature. We consider two key aspects, namely, high accuracy and low power. Applications considering high accuracy will benefit from choosing an NTN constellation using a large bandwidth signal such as StarLink, Kuiper, or OneWeb. These broadband NTN constellations use a very large bandwidth (100 to 250 MHz) and have a very high number of satellites in the constellation, leading to a very low GDOP [193]. For low-power applications where accuracy is not critical, such as asset tracking, IoT-focused NTNs can be considered. These networks use a narrow bandwidth and use a simpler modulation, requiring less power to process the signals. As a consequence, the data rate and localization accuracy will decrease. A combination of multiple constellations can also be considered. In [69], Iridium NEXT and Orbcomm satellites are used together as space anchors. The authors of [197] combine even four different constellations.
- 2) Signal Structure: When using SOPs for localization, knowing the basic signal structure is imperative. The most straightforward approach is to use NTNs that use standardized modulations. Several IoT-focused satellite constellations use NB-IoT. This standard is regulated by 3GPP, and in the recent Release 17, space-based NB-IoT networks were presented [205]. As a result, the signal structure is publicly known and could be exploited for localization. In addition, an altered version of LoRa is proposed for space-based networks, called long-range frequency hopping spread spectrum (LR-FHSS) [206], which will be used by the Lacuna constellation. Many other commercial communication satellite systems use proprietary signals which are not disclosed to the public. One strategy would be to reverse-engineer the signal structure.

TABLE VIII: Localization methods used in the literature using commercial LEO NTNs.

Name	Localization user-cases			
Name	TDOA	Doppler		
Iridium NEXT	[210]	[197][211][15]		
Starlink	-	[197][212][213][214]		
OneWeb	-	[197][213]		
GlobalStar	-	[15][214][215]		
Lacuna	-	[15]		
OrbComm	-	[197][211][15][214][216][217]		

This can be achieved by recording the signal and figuring out the signal parameters, for example, sampling frequency, modulation type, and frame structure. In [207], the authors recorded and analyzed StarLink signals. They described how they reverse-engineered the signals and documented the signal structure and synchronization sequences. They hope to exploit these sequences in the future as a means to measure the range to the receiver.

- 3) Ephemeris: In order to localize a target based on SOPs, the location of the anchors has to be known. The position of satellites is constantly changing. However, it is predictable as the satellite follows an orbit around Earth. Therefore, to calculate the position of the satellite, the orbit has to be known. The description of an orbit is called the ephemeris. The ephemeris can be presented by a two-line element (TLE). These TLEs are publicly shared by the NORAD (North American Aerospace Defense Command) and are updated daily [208]. TLEs might be an easily available source of ephemeris data; however, they are not very precise, which results in reduced positioning accuracy. In order to obtain precise ephemeris data, a reference network can be set up. Such a network of receivers tracks the satellites from fixed and known positions. Based on the measurements, the orbits of the satellites can be precisely determined, improving the positioning performance. In [209], the ephemeris data of LEO satellites is estimated by tracking the satellites using Doppler and pseudo-range measurements to improve upon the TLE data. A datalink is needed in order to transfer the ephemeris of the satellites to the target. This datalink can be a terrestrial link or provided by the NTN itself. Providing the data through a satellite link has the advantage of only having to receive one signal and reaching global coverage. Following 3GPP Rel. 17, space-born NB-IoT satellites will provide the ephemeris data to the users of the network [205].
- 4) Localization Methods: Localization using SOPs is a recent research domain that has attracted considerable attention. The recent addition of many new LEO communication satellites has led to a boom in research to use the new SOPs to localize terrestrial nodes.
- *Time-Difference-of-Arrival*: As this technique is very mature and well-researched, it is one of the prime candidates for being used with LEO SOPs. In order for this technique to work, the space anchors have to be accurately synchronized with each other, and the signal structure has to be known in order to use ranging techniques. GlobalStar is a prime candidate to be used as a basis for a SOP TDOA navigation

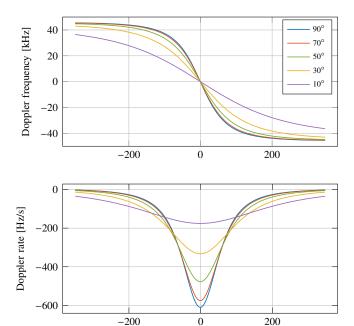


Fig. 11: The Doppler frequency and rate for passes of a LEO satellite with various maximum elevation angles $(10^{\circ}-90^{\circ})$. The orbit height was set to 550 km and the carrier frequency is 2 GHz. By tracking the Doppler frequency of a LEO satellite over a short period, the pseudorange rate can be determined. The location of the receiver can be calculated by tracking multiple satellites.

Time [s]

system as the GlobalStar signal structure is based on a code-division multiple access (CDMA) modulation similar to GNSS systems and has been studied before [218] [215].

• Doppler positioning: LEO satellites move across the sky rapidly. Due to the significant relative motion between the satellites and the ground-based targets, the signal experiences a significant Doppler shift. The basic idea is to track the Doppler frequency, which has a different profile for different satellites and depends on the location of the satellite in the sky. Fig. 11 shows the Doppler frequency and rate for LEO satellites with various maximum elevation angles. Depending on whether the satellite passes directly overhead (90°) or stays low at the horizon (0°), the maximum elevation angle of the satellite changes accordingly. The Doppler frequency relates to the relative speed of the satellite and the receiver. The values shown correspond to an orbit height of 550 km and a carrier frequency of 2 GHz, with a static receiver considered. When tracking the Doppler frequency of multiple anchors, a target can localize itself. This method was the basis of Transit, the first satellite-based localization system [48]. Moreover, the method is regaining popularity in research as a way to localize targets based on LEO SOPs. The main advantage compared to TDOA is that the exact signal structure is not needed to track the Doppler frequency of the signal. However, in order to reach a good positioning estimate, the signal has to be tracked for a longer time [69]. The Doppler frequency measurement can be converted to a pseudorange rate measurement as the Doppler frequency relates to the relative motion between the satellite and the target receiver. The pseudorange rate $z_n[k]$ at timestep k is calculated as

$$z_n[k] = c \frac{f_{D_n}[k]}{f_{c_n}},\tag{34}$$

where f_{D_n} is the Doppler frequency of satellite n, f_{c_n} is the carrier frequency of the n-th satellite [69], and c is the speed of light. When considering a static receiver, the pseudorange rate $\dot{\mathbf{p}}_{A,n}[k]$ relates to the location of the user \mathbf{p}_U and the satellite/anchor $\mathbf{p}_{A,n}[k]$ as

$$z_{n}[k] = \frac{\dot{\mathbf{p}}_{A,n}^{\mathsf{T}}[k] \left[\mathbf{p}_{U} - \mathbf{p}_{A,n}[k] \right]}{\|\mathbf{p}_{U} - \mathbf{p}_{A,n}[k]\|} + c \left(\dot{\delta}t_{U} - \dot{\delta}t_{A,n} \right) + \epsilon_{A,n}[k],$$
(35)

where δt_U and $\delta t_{A,n}$ are the clock drift of the receiver and the *n*-th satellite, respectively, and $\epsilon_{A,n}[k]$ includes errors induced by noise and atmospheric effects.

In [69], Doppler positioning is deployed on OrbComm and Iridium signals. The Doppler frequency and carrier phase are tracked using Costas Loops. The position of the user is estimated using an Extended Kalman Filter. In [197], Starlink and OneWeb satellites are included in the Doppler receiver design to further improve the achievable accuracy; they reach an accuracy of 5.8 m. In [213], the authors develop a point-solution using a batch least-squares filter to estimate position, velocity, clock offset, and clock offset rate by tracking eight or more satellites simultaneously. Their simulations and GDOP studies indicate that the possible accuracy is in the order of 1 to 5 meters.

Table VIII reveals that until now, most of the literature on positioning with LEO satellites is based on Doppler positioning. The main reason for this is that the exact signal structure is often unknown, and Doppler positioning does not require this knowledge.

C. Aerial Targets: Special Considerations

When using space anchors to localize aerial targets, some special considerations have to be taken into account, as detailed in the following.

- 1) Geometrical Aspects: As aerial users move around in 3D space, estimating the altitude of the aerial targets is of utmost importance. When using LEO satellites as anchors, the vertical GDOP tends to be double the Horizontal GDOP [189]. This leads to a halving of the accuracy when comparing the target's altitude estimation to its horizontal position estimation. The reason behind this poor vertical GDOP is mainly due to the fact that visible space anchors all have a positive elevation angle; therefore, the information of the vertical position only comes from above, while horizontal information comes from all four horizontal directions, i.e., north, south, west, and east.
- 2) Radio Aspects: Aerial targets experience a high chance of open-sky visibility compared to their ground counterparts, increasing the localization availability using space anchors. Aerial targets can make use of all satellites above the horizon, even the ones at low elevation, which are often unavailable for terrestrial targets. Moreover, aerial users are not typically in a rich multipath environment, lowering the impact of possible multipath errors [151]. Nevertheless, strong multipath effects can be present due to reflection from potential structural

elements of the aerial user, e.g., the wings of an airplane. The high visibility of aerial targets also makes them very prone to interference and jamming events [219][220]. This increases the need for interference mitigation as well as anti-jamming and anti-spoofing technologies.

3) Mobility Aspects: Aerial targets often exhibit a high mobility profile [147], which includes not only 3D movement but also changes in the 3D orientation of aerial targets. These movements can lead to a changing orientation of the receiving antenna. For instance, typical antennas used in current GNSS applications are optimized for receiving signals arriving from a positive elevation angle. In this way, the antennas focus on the signals coming from the sky while suppressing ground reflection and interference. However, when the orientation of an aerial target antenna changes due to the high 6D mobility, the quality of the received signal will degrade for certain space anchors as the antenna is directed away from them [220]. Furthermore, the received power of ground reflections and interference will increase.

D. Key Takeaways

Localization using LEO satellites represents a vibrant domain of research, propelled by the recent surge in LEO communication networks. A unique aspect of LEO satellites compared to ground and aerial anchors is their rapid movement in defined orbits. On the one hand, this movement unlocks Doppler-based localization and diversifies the set of satellites used in target localization, but on the other hand, it brings challenges related to synchronization, orbital errors, and dynamic resource allocation. Present research primarily focuses on harnessing LEO communication signals for localization purposes. However, these networks are not inherently tailored for positioning, necessitating receivers to opportunistically self-localize. Alternatively, dedicated LEO-PNT constellations are in developmental stages, exclusively engineered to provide precise and resilient positioning services. Future research should explore the synergy between these two approaches, combining communication capabilities with accurate positioning services within a unified NTN. Moreover, as LEO constellations continue to expand, commercial positioning services will become available, opening the door to more comprehensive studies on localization systems coexistence and integration within the space segment and across the multiple GAS segments.

VII. RESEARCH DIRECTIONS TOWARDS 6G GAS LOCALIZATION

With the wider bandwidth and larger number of antennas in 5G, the localization accuracy can attain meter-level [221], and in 6G, the localization accuracy is expected to realize centimeter-level. This section presents the role key 6G enablers are expected to play in GAS localization systems, exploring their added value and highlighting research directions towards 6G GAS localization.

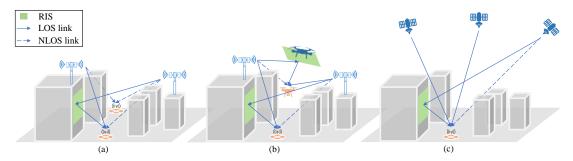


Fig. 12: A representation of RIS in GAS.

A. Reconfigurable Intelligent Surfaces

The main motivation behind RIS, for both positioning and communication purposes, aligns closely with that of integrated GAS networks, aiming to enhance the coverage, availability, and robustness of the solution. RIS offers a means to extend coverage by serving as supplementary anchors that can provide extra LOS positioning measurements, e.g., TOA, AOA, and AOD, for localization systems in 6G networks [26,222]. In cases where the LOS path between the UE and the TN/NTN is heavily degraded by the environment due to obstacles, the RIS alternative path can act as the main positioning path in the system. Furthermore, if the TN/NTN measurements were strong enough, then adding extra measurements to the fusion filter, in the form of RIS measurements, will enhance the localization accuracy further. It is worth noting that RIS measurements can be distinguished from other propagation paths by utilizing unique signatures, such as time encoding [223]. This is achieved by manipulating the phase or amplitude of the incident signals as they interact with the reconfigurable elements. These interactions are precisely controlled and can be finely tuned to create distinguishable temporal patterns in the signals reflected or refracted by the RIS.

RIS also introduces a distinctive advantage in high-precision positioning by providing an accurate estimation of horizontal and vertical AOA and AOD [25]. This is mainly achieved by performing a beam-sweep via the large number of reconfigurable elements spanning the RIS, as illustrated in Fig. 12a. Such accurate angular measurements can facilitate fine estimation of the position of a UE in a SISO system [224]. Hence, the requirement of having a 2D array at the UE, BS, or NTN for single anchor positioning can be relaxed in the presence of RIS. A noteworthy advantage of deploying the RIS technology, compared to deploying extra NTN/TN anchors, lies in its ability to improve positioning without introducing additional clock offsets or carrier offsets. Adding additional anchor nodes (terrestrial or non-terrestrial) to the system might complicate the processing chain due to synchronization challenges. On the other hand, RIS seamlessly integrates into existing systems without posing such hindrances. Additionally, the deployment and operation costs of RIS are considerably lower compared to TNs and NTNs [225]. This characteristic streamlines the deployment of RIS-enabled positioning solutions, making them more practical and efficient to implement in 6G networks.

Within the realm of RIS technology, various configurations emerge, including passive RIS, active RIS, hybrid RIS, and

simultaneous transmitting and reflecting (STAR)-RIS, also known as intelligent Omni-surface. In [24], the role of RIS in 6G localization has been comprehensively investigated in terms of different types of RISs. Passive RIS is designed to reflect the incident signals without power amplification. Hence, they are more energy efficient than the other implementations but can only operate if the incident signal is strong enough. Active RIS implementations, on the other hand, hold the potential to amplify the power of NTN signals, bolstering their efficacy for communications and localization purposes in degraded channel conditions [226]. As the name suggests, hybrid RIS comprises passive and active elements. Hybrid RIS can combine the simplicity and power efficiency of passive implementations while maintaining the favorable qualities of the active implementations [227]⁷. All of the aforementioned RIS implementations only consider reflecting the incident signal to any angle on the space in front of the RIS [228], which implies that both the user and the anchor node should be in front of the plane of the RIS to operate. STAR-RIS, on the other hand, proposed in [228], extends the operation of the RIS to the other side of the plane, enabling the signal to pass through the surface and be refracted to any angle on the other end. Hence, STAR-RIS presents an opportunity to extend the TN's and NTN's coverage to include indoor localization, effectively expanding the reach of the network's localization services. Yet, it is worth noting that increasing the reach of RIS may cause interference issues with other networks [229]. Hence, managing such interference in RISequipped networks may pose challenges, necessitating careful coordination strategies to ensure that the benefits of RIS are realized without introducing undue disruptions.

As presented in the localization measurables in Section III, the uncontrollable multipath nature of the wireless channel is one of the major localization error sources. In principle, one of the key advantages of RIS, from a localization perspective, is its ability to offer a controllable path in the channel model. The channel model between the transmitter and the receiver with a RIS found in between can be constructed by building upon the established model in (20):

$$\mathbf{y}_{k,l} = \mathbf{W}_{l}^{\mathsf{H}} (\mathbf{H}_{k,l} + \mathbf{H}_{k,l}') \mathbf{f}_{l} \mathbf{s}_{k,l} + \mathbf{n}_{k,l}, \qquad (36)$$

⁷The term *hybrid* RIS is sometimes used to indicate the usage of some active RIS elements that can sample the incoming signal via an RF chain [226]. In our case, however, *hybrid* RIS indicates the usage of some active elements that *actively* amplifies the impinging signal before reflection.

where $H_{k,l}$ and $H'_{k,l}$ comprise the uncontrollable and the semi-controllable RIS channels between the transmitter and the receiver, respectively. The semi-controllable channel constitutes the uncontrollable Tx-RIS $(H_{\text{tr}_{k,l}})$ and RIS-Rx $(H_{\text{ru}_{k,l}})$ channels and the controllable RIS response $\Omega_l = \text{diag}(\vartheta_l) \in \mathbb{C}^{N \times N}$, where N is the number of RIS elements, and $\vartheta_l \in \mathbb{C}^N$ is the RIS response vector. The elements of ϑ_l are controllable and are chosen from Θ , the set of feasible RIS element configurations, and are formulated as follows:

$$\boldsymbol{\vartheta}_l = \boldsymbol{\gamma}_l \odot \boldsymbol{\omega}_l, \tag{37}$$

where $\gamma_l \in \mathbb{R}^N$ constitutes the RIS elements' gain/attenuation factors, and $\omega_l = [e^{-J2\pi f_c \tau_{l,l}}, \dots, e^{-J2\pi f_c \tau_{N,l}}]^\top \in \mathbb{C}^N$ are the corresponding phase shifts. Here, $\tau_{n,l}$ is the delay caused by the n-th RIS element during the l-th symbol.

- 1) RIS-Aided Ground Anchors: Utilization of significantly large RIS setups with ground anchors offers the potential for near-field localization by harnessing wavefront curvature [230]. This capacity leverages the RIS's ability to manipulate signal propagation at extremely close distances, enabling advanced positioning techniques that capitalize on the unique behavior of electromagnetic waves in near-field scenarios. Thanks to RIS's ability to partially control multipath propagation [223], it is expected to have a positive impact on all localization KPIs when used with ground anchors. Key research challenges when using RIS with ground anchors are related to the operation of the RIS itself, e.g., nonlinearity and mutual coupling [223], and appropriate site selection, service access [231].
- 2) RIS-Aided Aerial Anchors: To overcome challenges related to RIS site selection, aerial deployment of RIS is explored in [231]. Deploying RISs on aerial anchors, as depicted in Fig. 12b, is mainly driven by the extra DOF obtained from the flexibility of anchors placement, which offers even more possibilities for LOS links with aerial and ground users, not only outdoor but also indoor [232]. Moreover, given the limited payload of aerial anchors, RIS can act as an alternative to carrying heavy RF transceivers required for coverage extension, positively impacting the UAV mission duration. Key challenges associated with RIS-aided aerial anchors mainly relate to the complexity of the localization system. In particular, this complexity arises from coordinating interference, incorporating RISs with antennas needed for flying assets in a compact and energy-efficient way, and implementing effective controllers for the RIS configuration given that the channel might be changing aggressively [233]. Moreover, considering the mobile nature of aerial anchors and their placement inaccuracy (cf. Section V-A4), errors in the RIS placement and geometric layout are unavoidable in practice, necessitating the localization and calibration of the RIS itself [46].
- 3) RIS-Aided Space Anchors: RISs can aid space anchors by deploying them on the ground as depicted in Fig. 12c and as explored in [234]. Integrating RISs on space anchors is investigated in [235], where they are exploited to aid cooperation for inter-satellite THz links in the LEO constellation. While such

implementation may indirectly aid several localization KPIs, its scalability and complexity require further investigation.

B. Joint Communication, Localization, and Sensing

To boost system integration and suitability in 6G, joint consideration of communication, localization, and sensing functionalities becomes prevailing. A comprehensive study on joint communication, localization, and sensing integrated in 6G radio technologies classifies them into levels. These levels are defined for communication as very short range (C1) and short-range (C2), for localization as high-accuracy (L1), lowlatency (L2), and low-complexity (L3), and for sensing as monostatic (S1), and Bi-multi-static (S2). The requirements to integrate them are discussed in terms of signals, hardware architectures, and deployments. In addition to the integration, the assistance/enhancement among different functions plays a vital role in harnessing the added value of such integration, which spawns a variety of research topics, such as, among others, sensing-aided communications and localizationaided communications. In terms of multilateral assistance, the research challenges, as well as potential directions, can be summarized as follows.

- Communication-aided sensing/localization: The relevantly large communication bandwidth, e.g., FR1, FR2, and the expected GHz scale in future 6G, plays a key role in terms of delay/range resolvability and accuracy of both localization and sensing [236]. Moreover, since communication networks are deployed aiming for wide coverage in JCAS⁸ systems, such coverage can also serve localization and sensing coverage. In addition, instead of depending solely on localization and sounding signals, e.g., PRS and SRS, anchors can jointly use communication signals for localization and sensing, which positively impacts localization latency, eliminating the localization latency bounds imposed by the period at which such signals are broadcast. As a targeted goal of JCAS, sharing communication hardware in transceivers also for localization and sensing benefits the power efficiency of the system [10].
- Localization-aided communication/sensing: Communication systems are tied to location information in various ways, including distances, delays, velocities, angles, and predictable user behavior [86]. With accurate and ubiquitous location information, communication performance can be enhanced across all layers of the communication protocol stack, such as indicating SNR and shadowing autocorrelation, predicting user mobility patterns, and securing driving operations. Moreover, wide-sense sensing naturally includes the positioning process. However, for multidimensional sensing, angular and Doppler information is needed, where localization can somehow increase sensing ability and reduce sensing complexity at the same time. In scenarios where narrow beams are used for communications, accurate knowledge of user location can aid in minimizing the number of adjacent

⁸We use the acronym JCAS to stay in line with the standard acronym used in the literature. Nonetheless, in this work, we stress that the sensing part of the acronym refers to the broad definition of sensing that includes passive sensing and active target localization.

- beams employed for environment sensing and blockage prediction [237].
- Sensing-aided localization/communication: Sensing, in the context of environment passive sensing, is generally considered together with the localization, e.g., SLAM technique [14]. The positioning of the target can benefit from accurate environment mapping via static background removal. Sensing-aided communication has been under extensive study in recent years, where the realization of joint communication and sensing has been proposed and tested in spatial, temporal, and frequency domains [10].

Towards practical implementation of JCAS systems, the potential topics proposed in [238] for JCAS systems include the following aspects: 1) Angle, range, and Doppler resolvability when employing high frequency such as mmWave and THz, 2) Directional beamforming and context awareness, 3) Joint waveforms and joint hardware, and 4) Algorithmic developments including model-based MIMO algorithms and data-driven AI. These general topics are also promising in the JCAS system. However, in the following, we present tailored JCAS topics for the various GAS anchors.

- 1) JCAS-Aided Ground Anchors: Considering JCAS systems in 6G, techniques such as SLAM can be enabled by relying on communication signals. To this end, the waveform design of communication signals to jointly allow accurate SLAM is an important research direction. By employing JCAS in ground anchors, the localization performance can be boosted by identifying the scatterers in the environment, where a specific multipath can serve as a virtual anchor. The study in [221] shows through learning the locations of such virtual anchors, i.e., scatterers, the localization performance can achieve an accuracy of 10 cm, outperforming the 60 cm accuracy obtained when using localization systems solely. Thanks to the rich scattering in ground networks, multipath components serving as virtual anchors can help improve the localization accuracy that is constrained by the number of physical anchors.
- 2) JCAS-Aided Aerial Anchors: Thanks to the umbrella view of aerial anchors on a given area, they offer the option to fuse RF-based measurements, such as CSI, with vision-based sensing data captured from the equipped cameras. Under this context, a promising application resides in environmental reconstruction, which includes both static scatterers and mobile targets such as cars and humans. Thus, a data-fusion-based environment mapping is able to provide accurate location information of moving targets by using a swarm to form spatial diversity [239] or a single moving aircraft to collect time-series information. Moreover, aerial anchors can be used in conjunction with ground anchors to form a bistatic JCAS system, where the aerial anchor collects a ground BS's echoed signals for radar sensing [240].
- 3) JCAS-Aided Space Anchors: Jointly considering sensing/radar and communication in LEO satellites is a rising trend. Recent works such as [241] optimized the hybrid precoding to balance the radar beam pattern and energy efficiency of communication in a massive MIMO-enabled LEO satellite. It provides a solution to use sub-arrays to sense targets and serve users simultaneously. Besides, the

LOS-dominant channel between space anchors facilitates the exchange of information free of multipath interference, e.g., via THz narrow beams, which suggests space anchors can be easily synchronized, thus positively impacting localization accuracy and resource allocation in general.

C. AI-Empowered Localization

AI refers to a collection of algorithms and models employed in devices to enable them to cognitively plan and make optimum decisions. To train such algorithms and models, AI systems rely on ML, which generates versatile models by exploiting relatively large amounts of data for model training [242,243]. This model training can be done in a supervised, unsupervised, semi-supervised, or reinforcement learning manner. AI-aided localization can be broadly categorized into two forms: A fully data-driven form and a complementary AI-enhanced model-driven form.

- Data-driven: In this form, AI-based algorithms are designed, taking raw measurements and RFs fingerprints, which may include power, time, and/or angle, to provide a position estimate [5,244]. Data-driven learning, in which data-trained models replace the full location estimation functionality, can potentially be beneficial in terms of computational complexity during the location estimation/inference phase [5]. However, the main drawback is the need for very large and representative datasets [245,246]. In order to reduce reliance on location-based labeled datasets, deep-reinforcementlearning-based unsupervised localization methods can be employed [246]. When the statistics of the true and trained channels are not the same, blind data-driven methods can fail dramatically; in such cases, online learning is a promising solution at the cost of computational complexity and localization latency. Furthermore, the data-driven localization using measurements collected at the receiver can be extended to end-to-end learning for localization, jointly optimizing transmit beamformers and receiver-side algorithms, even in the presence of hardware impairments [247].
- AI-enhanced model-driven: In this form, AI-based algorithms are designed to model the stochastic propagation environment, e.g., channel models [248], or to calibrate and compensate for imperfections in localization measurables [30]. One promising direction is the calibration in antenna arrays, e.g., ML-MUSIC, where ML is used to calibrate the array imperfections [249]. In addition, clock offset and hardware imperfections that make localization challenging are also relevant to AI-empowered localization [30]. In addition, AI could be used to learn how to fuse multiple estimates in a weighted manner or make a final decision from noisy estimates obtained via different anchors in the GAS network.

Intuitively, AI-enhanced model-driven methods that tackle calibration and hardware imperfections are beneficial for all GAS anchors, whereas the fully data-driven methods are more beneficial for ground and LAP anchors compared to HAP and space ones. In the following, we discuss the AI-empowered localization from each of the GAS anchors' perspectives.

- 1) AI-Aided Ground Anchors: Both data-driven AI and AI-enhanced model-driven approaches can be exploited in scenarios with ground anchors. In fully data-driven methods, the rich multipath nature of the G2G environments ensure sufficient variations in datasets features captured in the spatial domain, e.g., power, time, and angle features [5,244]. In G2A scenarios, where the multipath effect is less pronounced, AIempowered data-driven solutions can rely on feature variation in the spatial domain over relatively wide areas, e.g., tracking aircraft over ranges of several kilometers, at the cost of limited localization accuracy [146]. In AI-enhanced modeldriven methods, ML classifiers can be used to identify the LOS and NLOS signals received at the anchor, boosting the accuracy of time-based localization methods [250]. MLbased clock-drift modeling and offset compensation has been explored in [30], enabling TDOA-based localization in widely distributed ground anchors.
- 2) AI-Aided Aerial Anchors: LAP aerial anchors can exploit data-driven methods in A2G localization, benefiting from the rich multipath ground environments and the mixed LOS/NLOS links compared to the predominantly LOS links in the case of HAPs. In addition to the calibration and channel modeling advantages that can be obtained from AI-enhanced model-driven methods [251], 3D placement and trajectory design of aerial anchors can be optimized using reinforcement learning [37]. A critical issue that requires careful attention when employing ML models in aerial anchors is the computational complexity and the corresponding power consumption of model training with large datasets. To address this issue, efficient methods for online learning are needed in order to work in resource-constrained aerial anchors [251]. Moreover, federated learning is investigated in the literature to address the resource-constrained problem by distributing the computations load; however, it does not fully address the communication bottleneck, which is still an open research area, since AI models' weights still need to be transmitted [252].
- 3) AI-Aided Space Anchors: The majority of works focusing on AI-aided space anchors address the resource allocation problem due to the need to optimize the scarce and expensive satellite resources [7]. In data-driven-based localization, several works investigated using LEO satellite images in 2D [253] or 3D [129], aiding ground anchors in keeping up-to-date environment maps. Due to the relatively high speed of space anchors in the LEO orbit, space anchors can benefit from AI-enhanced model-driven methods, which enable orbit and clock/Doppler offset patterns prediction, improving target tracking accuracy [254].

D. Cell-Free Paradigm

Current 5G networks address the demand for higher spectral efficiency through densifying BS (a.k.a. AP) deployment with small cells and increasing the number of antennas per base station, i.e., massive MIMO. While this network-centric approach noticeably increases spectral efficiency, it makes the intercell interference problem more pronounced, which particularly impacts cell-edge users, e.g., they might be tens of dB weaker channel compared to cell-center users [255,256]. Intuitively,

a given UE would only be affected by interference between its own cell and a set of neighboring cells; hence, it is only this corresponding cluster of APs that needs to cooperate to alleviate inter-cell interference for this UE. This intuition is the foundation of the emerging user-centric CF paradigm, which adopts an alternative network architecture to address the inadequate cell-edge users' experience. In CF networks, a user-centric cluster of APs jointly operates to coherently serve UEs on the same time/frequency resource, using spatial multiplexing. In order to have sufficient spatial multiplexing capabilities, a massive number of distributed APs are needed, introducing cell-free massive MIMO (CF-mMIMO), in which the word *massive* implies more APs than UEs [255]. The architecture of CF-mMIMO inherently implies (sub-)nanosecond synchronization among densely deployed distributed APs, and potentially multiple antennas both in uplink and downlink, which benefit from reduced pathloss and macro-diversity mitigating shadowing/blocking [257]. Having a massive number of APs, each acting as a localization anchor, as visualized in Fig. 13, offers unprecedented opportunities to realize a robust GDOP across a wide area, positively impacting localization accuracy and coverage in all segments of GAS networks. In addition, CF as a network architecture could be a potential candidate to realize spatial multiplexing using the different segments of GAS networks since LOS MIMO essentially requires a large antenna separation. The implementation of the cell-free architecture can take various forms in terms of using several APs with a single or multiple antennas. The coherent transmission in cell-free networks can be done in time, phase, or both [256,258,259], whichever used directly influenced the localization system design.

- Time-coherent operation: Operating in a time-coherent manner can be done using several single-antenna APs that are synchronized by a GPS clock. Such tight synchronization provides accurate TOA measurements with respect to each other, enabling target localization using TDOA or RTT [30].
- Phase-coherent operation: A more popular form of cell-free is the phase-coherent transmission [255], where phases among APs are stable during each transmission period. In this form, APs are likely to have spatially distributed multiple antennas [255]. From a localization perspective, this form of cell-free gives both time and phase information across the spatially distributed arrays.

Despite the enormous advantages of CF-mMIMO, challenges related to synchronizing dynamic APs clusters and resource allocation are open research questions [255,256]. In the following, we present CF potential impact within GAS anchors.

1) Cell-Free-Aided Ground Anchors: Beyond communication, CF networks have also been shown advantageous for sensing [260] and localization [249] using ground anchors. The essential benefit is the extra DOF obtained from the relatively large number of cooperative distributed anchors, i.e., APs, improving GDOP of location estimates. The rather large number of APs envisioned in CF networks enables them to serve ground and aerial users jointly [261]. Cell-free networks could also be envisioned to provide localization

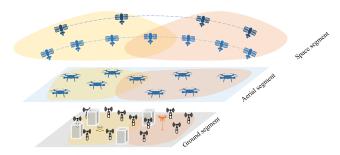


Fig. 13: A representation of cell-free in the different segments of GAS networks.

of not only ground and aerial active targets but also passive ones; however, a detailed investigation is needed to understand whether the benefits outweigh the drawbacks from energy and computational perspectives.

- 2) Cell-Free Aided Aerial Anchors: Similar to ground anchors, employing a CF network architecture using aerial anchors benefits localization performance, offering extra DOFs. In general, synchronization is a critical challenge in CF networks [256], limiting the accuracy of time-based localization methods. On the one hand, aerial anchors can exploit the high probability of LOS links to employ over-the-air synchronization. However, on the other hand, the mobile nature of aerial anchors leads to even more dynamic AP clusters, necessitating frequent synchronization. This same mobile nature of aerial anchors can be exploited to optimize the aerial APs placement in a UAV-based CF to serve mobile UE [262,263].
- 3) Cell-Free-Aided Space Anchors: The concept of building distributed antenna arrays for the reception of signals from space is essentially used in radio astronomy as a powerful method to increase RSS [264]. While these deployments use distributed ground stations, the concept could be extended to the aerial and space segments if anchor density increases there. In such cases, the spatial multiplexing needs to be done in 3D, e.g., by having multisegment anchor cooperation [265], which in turn also serves target localization in 3D. CF-mMIMO using ultra-dense LEO satellite constellations is introduced in [266]. However, the authors addressed the network architecture only from pilot assignment, beamforming, and handover management. The localization problem using CF-mMIMO using ultra-dense LEO satellite constellations is an open research question.

E. THz Band

THz communications are considered as a 6G enabler, going beyond mmWave systems, covering communications in the band of 0.1-10 THz [267]. Moving from mmWave in 5G to THz 6G systems, higher frequencies, larger bandwidths, and larger array sizes with smaller footprints are expected. The high frequencies at the THz band increase pathloss, reduce multipath components, and the corresponding high bandwidths provide relatively high path delay resolvability. These characteristics make THz localization most fit for ground anchors. In [11], the authors presented a detailed tutorial on radio localization in the THz band, where mainly ground anchors are explored. A comparison of localization merits and challenges

between mmWave and THz is provided. This comparison includes several aspects, such as array size, array type, hardware imperfections, synchronization, and propagation effect. In the following, we discuss THz from the GAS anchors' perspective.

- 1) THz-Aided Ground Anchors: Existing works on THz localization address system structures, localization algorithms, and simulation platforms. In particular, AOA-based localization is attracting considerable research focus due to the narrow beams used in THz that result in a very high angle resolvability [11]. Moreover, exploiting RIS and THz is a promising research direction, enabling near-field localization and channel estimation [268].
- 2) THz-Aided Aerial Anchors: Similar to ground anchors, THz can boost the AOA-based localization using LAP aerial anchors. However, the difficulties that aerial anchors face in maintaining their position, e.g., wobbling because of wind, can lead to frequent beam misalignment, negatively influencing the localization stability [269]. In addition, THz can aid UAV-to-UAV communications, ensuring low overhead resources and interference management.
- 3) THz-Aided Space Anchors: In space anchors, THz is mainly explored for inter-LEO satellite communications [235], which may indirectly benefit the localization performance of ground and aerial anchors, e.g., by improving resource allocation and interference management between space anchors.

F. 6G KVIs

The 6G enablers discussed in this section introduce new DOFs and design aspects, promising applications with unprecedented demands in terms of localization KPIs such as extended reality and digital twin [270]. KVIs are adopted in 6G networks in order to broaden the assessment of future wireless networks beyond the technical aspects-focused KPIs, to also include forward-looking societal aspects [271]. They aim to better capture the spirit of the sustainable development goals defined by the United Nations [272]. In the following, we present vital KVIs related to localization in GAS networks.

1) Trustworthiness: Localization security, tegrity/robustness, and user privacy can be embedded in the trustworthiness KVI [271]. In GAS localization systems, where the distance between the anchors and targets is relatively very large, e.g., as in the case of aerial and space anchors, localization signals are more prone to jamming and spoofing [273,274]. While maintaining the localization system's integrity is important to ground targets, its degradation typically has a broader impact on the aerial targets (due to lower shadowing), and it could lead to serious or even life-threatening consequences. Due to the many recently proposed and/or launched LEO constellations, a huge opportunity to strengthen our PNT-dependent infrastructure arises. The various radio signals from GAS anchors can be used to increase the diversity gain, boosting localization system robustness by exploiting redundancy to detect and eliminate faults. Moreover, with the large number of GAS anchors, jamming and spoofing attacks can be partially mitigated by detecting attacks and switching to anchors or systems that are not jammed or spoofed [274]. Despite this opportunity, when using SOPs from NTN anchors, there is no guarantee that the used signal is always correct and consistent. Any lack of consistency in the used SOP might impede the navigation service, which is particularly critical for aerial users as guaranteeing a position fix at all times is essential. To combat this shortcoming, cooperation with the constellation operator can be considered. This concept, called fused LEO GNSS [275], requires only a small part of the downlink capacity of modern mega-constellations to provide reliable PNT services with a higher accuracy than traditional GNSS. In addition, using LEO's SOPs in conjunction with another navigation system, e.g., GNSS, LEO-PNT, aerial-based or ground-based systems, will strengthen both systems' robustness and security flaws detection.

- 2) Sustainability: Wireless networks' sustainability includes energy efficiency, reusability, and maintenance. There are two aspects of sustainability in 6G GAS localization systems, namely, using localization to ensure sustainable communications and making the localization process sustainable in itself. In the former, accurate user location information facilitates a more efficient resource allocation, e.g., energyefficient and swift beam alignment [9], which is critical for battery-limited aerial anchors. In the latter, accurate localization systems should be designed to meet KPIs, but with constraints on allocated resources and energy consumption. Relevant examples include energy-constrained trajectory design for aerial anchors and the number of LEO satellites allocated to localize a target. Regarding reusability, JCAS is expected to play a major role by designing waveforms and hardware that support multi-functionalities.
- 3) Inclusiveness: While the localization coverage as a KPI can be used to assess the geographical area in which accurate localization is available, inclusiveness goes beyond coverage to also include the accessibility and affordability of the localization system [271]. GAS-based localization, mainly using aerial and space anchors, is expected to have a significant impact on the localization inclusiveness aspect by offering global connectivity and localization coverage using off-the-shelf equipment [214].
- 4) Harmonized Operation: In Release-17 of 3GPP, two new FR1 bands for NTN are included, namely, NTN 1.6 GHz and n256 NTN 2 GHz. Several additional frequencies within those ranges are being discussed within 3GPP for additional NTN spectrum [21]. Moreover, in the long term, FR3 band (7-24 GHz) is emerging [276], along with bands above 10 GHz are targeted as well (around 17 GHz for UL and 27 GHz for DL), offering more frequency options to use for improving localization performance. As the spectrum gets more crowded, it becomes more difficult to avoid interference. An interesting and important topic for research is how we could make GAS localization networks interference-resilient. Given the high output power requirements in GAS networks, it becomes difficult to meet tight spectrum mask requirements. Current ongoing research on out-of-band radiation for Massive MIMO and CF-mMIMO is also of interest here, and an important topic [277], as out-of-band radiation also gets beamformed, especially in a LOS scenario, which occurs in high probability in aerial and space segments.

G. Key Takeaways

6G enablers, which include RIS, JCAS, AI, CF, and THz are expected to play important roles in pushing localization system KPIs to unprecedented levels. A great amount of research effort was put into studying RIS-aided localization systems, where the majority of works focus on using RIS with ground anchors. Investigating RIS integration with aerial or space anchors is an interesting open research question, where relatively long distances and high anchor mobility are open challenges. JCAS is a 6G enabler that offers a unified costefficient communication, sensing, and localization framework. While ongoing research efforts on JCAS focus on waveform design and hardware architecture, the network architecture, and deployment scenarios when considering single or multiple GAS segments are open research problems. Regarding network architecture, CF, as a potential architecture in 6G networks, can boost the localization performance by exploiting the massive deployment of APs acting as anchors. However, time and phase synchronization, as well as resource allocation, are open research questions in CF networks, even in communication use cases [257,278]. To tackle synchronization and calibration challenges in wireless networks, AI-empowered solutions are gaining considerable attention. AI-empowered model-driven localization solutions can help address calibration and imperfections at all 3 GAS anchors; however, more effort needs to be invested in obtaining labeled datasets, which is a challenging task, particularly for aerial and space anchors. Another 6G enabler is the THz band. THz localization solutions enable remarkably accurate time and angle measurements. However, due to the relatively high pathloss, the use of THz in NTNs will potentially focus on inter-anchor links, e.g., for synchronization and measurements fusion. Finally, KVIs, such as trustworthiness and sustainability, are envisioned to complement existing KPIs, opening new performance evaluation aspects.

VIII. CONCLUSION

In this tutorial paper, we addressed the localization problem in GAS networks. Localization fundamentals, considering the 3D space of GAS networks, have been detailed, including coordination systems, localization measurables, methods, and KPIs. Moreover, we presented the system model, design aspects, and special considerations for ground, aerial, and space anchors, covering both ground and aerial targets' localization. The promising potential and advantages of using each of the GAS anchors have been explored, and key results have been discussed, giving a quantified perspective. We explored the role of the main 6G enablers, namely, RIS, JCAS, CF-mMIMO, AI, and THz, in the localization problem in GAS networks, highlighting the prospective impact these enablers have on the different localization aspects, KPIs, and KVIs.

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